

REMARKS

Favorable consideration of the pending claims is respectfully requested. Claims 1 and 5 have been amended and claims 8-13 have been added. Claims 1-13 are pending following entry of this amendment, with Claims 1, 5 and 9 being in independent format.

Claim 1 has been amended to remove alternative claim language. The deleted subject matter has been claimed in newly added claims 9 and 13. Claim 5 has been amended to remove some alternative claim language, and the subject matter deleted from claim 5 is recited in new claim 8. New claims 10-12 recite subject matter previously recited in claims 2-4 with respect to the subject matter deleted from claim 1 and presented in independent claim 9. It is urged that these claim amendments are supported by the specification as it was originally filed.

The undersigned Applicant's representative was given Power of Attorney in this application following a change in the assignee of this application. The undersigned reviewed this application, its file history and the original file materials and notes that the specification, claims and abstract, as filed, included 76 pages. The specification provided on the USPTO PAIR site as the filed specification is correct, but the specification published as US 2004/0254461 A1, although it includes the drawings as filed and the claims as filed, omits numerous pages of the specification as filed.

The specification as filed was in an unusual format. It included five (5) pages, including a cross-reference to related applications, a Summary, a Brief Description of Drawings 24A-26, and a Detailed Description of the Invention, which was followed by a copy of the parent PCT application published as PCT/US00/14691, starting with the Summary of the Invention. This is an unusual format for filing a continuation-in-part patent application, but to the knowledge of the undersigned, it is acceptable and the full specification, including all of the pages as filed, constitutes the specification as filed. The Office has requested that we file a Substitute Specification conforming unusual specifications to the usual format in previous cases, and we have therefore amended the specification, as filed, to conform to the usual U.S. patent application format.

We herewith submit an amended and Substitute Specification, in which we have amended the specification, as filed, to conform to US practice. The reference to priority applications has been amended and updated. We have rearranged the description provided in the first several

pages of the specification as filed and integrated that description into the body of the “main” portion of the specification. This is simply a rearrangement of the previously filed text; no new matter has been added. Grammatical and typographical errors have been corrected, where noted, in the amended specification. The Substitute Specification is submitted in both a marked-up version and a clean version. It is urged that no new matter has been introduced into the substitute specification, and entry of the substitute specification is requested.

Applicants do not believe that it is necessary to republish the application, as filed, and therefore have not submitted a Request for Republication. Applicants do request, however, that the substitute specification be used as the basis for publishing the issued U.S. patent when claims in this application have been allowed.

Claim rejections under 35 USC §103

Claims 1-7 stand rejected under 35 USC §103(a) as being unpatentable over the combined teachings of Dodd et al., U.S. Patent 5,833,614 in view of Haider, U.S. Patent 6,135,963. This rejection is respectfully traversed.

Dodd et al. disclose an improved method for harmonic imaging that involves transmitting ultrasound at a fundamental frequency and receiving reflected ultrasound at a harmonic of the fundamental frequency. Dodd et al. disclose, as noted by the Examiner, that the transmission burst may be pulse width modified to reduce the harmonic response. The modulated waveform of Dodd et al. is also *amplitude* modified, whereas applicants’ claims specify that *the pulse amplitude is constant from pulse to pulse across the probe*. Applicants also don’t perceive that Dodd et al. discloses varying the number of pulses per burst while keeping each pulse amplitude constant from pulse to pulse across the probe, as specified in applicants’ independent claim 9.

Haider is cited for a showing of using pulse width modulation with less energy in a uniform progression to the array edge. Applicants submit, however, that Haider does not overcome the deficiencies of Dodd et al. with respect to applicants’ independent claims. It is urged, therefore, that there is no combination of Dodd et al. with Haider that would render applicants’ pending claims obvious in the manner required by 35 U.S.C. 103(a).


Concluding Remarks

Favorable reconsideration and early allowance of pending claims 1-13 is respectfully requested.

Charge Deposit Account

Please charge any additional fees that may be required, or credit any overpayment, to our Deposit Account No. 19-3555.

Respectfully submitted,

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Date: August 16, 2007

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20601

~~IN THE UNITED STATES PATENT AND TRADEMARK OFFICE~~
~~PATENT APPLICATION~~

**Title:—ACOUSTIC BEAM SHAPING BY PULSE POWER MODULATION AT
CONSTANT AMPLITUDE**

Inventor:—William Ackerman

Priority:—CROSS-REFERENCE TO RELATED APPLICATIONS:

This application is a ~~CIP~~continuation-in-part of US Application #No. 09/926,666, filed
~~November 30, 2001~~March 20, 2002, and ~~scheduled to issued on~~ January 27, 2004 as US Patent
#No. 6,682,483 and depends for priority on US Provisional Application #60/446,162, filed February
10, 2003. Application #09/926,666 depends for priority on international application which was a US
National Phase Application of PCT/US00/14691, filed May 26, 2000, which claims priority to—The
International application depends for priority on US Provisional Applications #Nos. 60/136,364
filed May 28, 1999, #No. 60/138,793 filed June 14, 1999 and #No. 60/152,886 filed September 8,
1999. This application also claims priority to U.S. Provisional Application No. 60/466,162 filed
February 10, 2003.

BACKGROUND OF THE INVENTION:

1.—Field of the Invention:

This invention relates to ultrasonic acoustic imaging, primarily for medical purposes.

2.—Brief Description of the Background Art.

_____ Ultrasonic acoustic imaging finds many uses, particularly in the field of non-invasive

medical testing. Direct detection of the emitted acoustic frequency permits, for example, prenatal fetal imaging. Detection of Doppler shifted acoustic frequencies permits observation of flow of a particle-containing liquid, for example, blood flow. Acoustic imaging equipment utilizes a probe that is applied to the skin of the patient overlying the part of the body being investigated. At the end of the probe is an array of transducers, usually piezoelectric, that are excited by bursts of electrical energy at the ultrasonic frequency and, for example, square wave modulated to transmit square wave modulated bursts of ultrasonic energy into the body region being investigated. The subsurface structures reflect some of that energy, either at the transmitted frequency or Doppler shifted, back to the probe, where it is detected by piezoelectric receiver elements in the probe.

One application of this technology to the three dimensional mapping and tracking of blood flow is

disclosed in parent US Application #No. 09/926,666, which is ~~scheduled to~~ issued on January 27, 2004 as US Patent #No. 6,682,483. The pertinent text of that application is included herein below.

~~SUMMARY OF THE INVENTION:~~

~~In traditional ultrasound applications the transmit signal is a short series of constant amplitude pulses. The pulses are 50 percent duty cycle pulse of acoustic energy at a single frequency. The amplitude must be very consistent or system performance will suffer. More advanced ultrasound techniques require shaping the transmitted beam by applying different power levels to different elements of an array of transmitter elements. However, the need to change voltage from one series of pulses to the next can cause problems with the electronics associated with the transmission. One problem is there is not enough time for the voltage to settle from one series to the next. This has the effect of causing artifacts in the image.~~

One means of overcoming the problems is to hold the voltage constant and modify a series of pulses to achieve the different power levels. Two techniques can be used with the pulses to control the power. The first is using fewer pulses in the series and the second is modifying the width of the pulses, thus, pulse width modulation. When using fewer pulses, the start of the series is delayed to position the center of series in the correct relationship with the other series. This invention employs both techniques. By varying both number and width a wide range of power levels can be achieved. The bandwidth characteristic of the piezoelectric element and system smoothes the pulse width modulated signal into a lower amplitude signal with greatly reduced harmonics. A second benefit is the reduced complexity of the transmit circuitry. Without this invention a significant amount of high power low impedance circuitry is required to rapidly change and hold the voltage at a new level. To do this with precision adds even more circuitry. The amount of digital logic necessary to create this new pulse width modulated transmit signal is small and does not significantly affect the cost of the system.

Brief Description of the Drawings

Figure 24A illustrates a series of control signals for transmitting a 50 percent duty cycle burst of ultrasonic energy at a reference amplitude.

Figure 24B illustrates a series of control signals that implement the prior art method of reducing the burst of energy by reducing the amplitude of the control signal.

Figure 24C illustrates a series of control signals that implement the herein disclosed method of reducing the burst energy by reducing the width of the pulses to significantly less than 50 percent

duty cycle.

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~~Figure 24D illustrates a series of control signals that product a reduction of bursts of energy by also reducing the number of pulses in the burst, while keeping the center of the burst in correct temporal relationship with other bursts.~~

~~Figure 25 is a circuit diagram of an exemplary apparatus for controlling and transmitting a series of bursts of varying amplitude.~~

~~Figure 26 is a circuit diagram of an exemplary transmission apparatus for varying burst of power by controlling pulse width and/or number of pluses per burst.~~

~~DETAILED DESCRIPTION OF THE INVENTION~~

~~Figures 24A-24D represent four different excitations of one transmitter transducer element of an ultrasonic probe. Each square wave (a, b, c, d) represents a constant amplitude pulse of ultrasonic excitation of defined pulse width. The series of five pulses illustrated in Fig. 24A will transmit a burst of acoustic energy into the patient at a defined power level. It has been found that, in order to reduce the level of side lobe transmission, it is desirable to excite probe elements at the edges of the probe (See, for example, Fig. 2) at a lower power level than elements at the center of the probe. In the prior art this is done by reducing the amplitude of transducer element excitation, as illustrated in Fig. 24B.~~

~~In the inventive method described herein, the transmitted power is reduced by reducing the pulse width, as illustrated in Fig. 24C or by reducing the number of pulses in each burst, as illustrated in Fig. 24D. The reduced number of pulses can be at a reduced probe width to achieve~~

~~a wide range of power level control, as illustrated in Fig. 24D, or at the full pulse width, which may be easier to implement electronically.~~

~~The amount of power reduction at the probe periphery should be sufficient to produce an operationally significant reduction of harmonics in the modulated signal.~~

~~Figure 25 illustrates exemplary circuitry used to excite probe elements at variable amplitude and pulse count. Figure 26 illustrates the much simpler circuitry with which probe elements are excited at constant amplitude with variable pulse width and variable number of pulses.~~

~~The disclosure that follows illustrates the exemplary use of Doppler shifted acoustic imaging to map and track blood flow. It illustrates the techniques and systems that profit from application of the hereindisclosed invention. Alternative probe geometries to which the hereindisclosed invention could be profitably applied are illustrated in US Pat. #6,524,253 B1, which is incorporated by reference in its entirety herein.~~

SUMMARY OF THE INVENTION

There is provided, in accordance with the present invention, a new, useful, and
10 unobvious method of determining parameters of blood flow, such as vector velocity,
blood flow volume, and Doppler spectral distribution, using sonic energy (ultrasound)
and a novel thinned array. Also provided is a novel method of tracking blood flow and
generating a three dimensional image of a blood vessel of interest that has much greater
resolution than images produced using heretofore known ultrasound devices
15 and methods.

Broadly, the present invention extends to a method for determining a parameter of blood
flow in a blood vessel of interest, comprising the steps of:

- 20 a) providing an array of sonic transducer elements, wherein the element spacing in
the array is greater than, equal or less than a half wavelength of the sonic energy
produced by the elements, wherein at least one element transmits sonic energy, and a
portion of the elements receive sonic energy;
- 25 b) directing sonic energy produced by the at least one element of the array into a
volume of the subject's body having the blood vessel of interest,
- c) receiving echoes of the sonic energy from the volume of the subject's body
having the blood vessel of interest;
- 30 d) reporting the echoes to a processor programmed to

- 35
- i) Doppler process the echoes to determine radial velocity of the blood flowing in the blood vessel of interest;
 - ii) calculate a three dimensional position of blood flow in the vessel of interest; and

- iii) calculate the parameter of blood flow in the blood vessel at the three dimensional position calculated in step (ii); and

- 5 (e) displaying the parameter on a display monitor that is electrically connected to 5 the processor.

Moreover, a method of the present invention permits an operator examining a subject to obtain information on blood flow in a particular region of the blood vessel of interest.

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As used herein, the phrases “element spacing” and “distance between the elements” can be used interchangeably and refer to the distance between the center of elements of an array.

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Various methods can be used to determine the three dimensional position of blood flow. In a particular embodiment, the method comprises the steps of having the processor programmed to:

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- i) determine a sum beam, an azimuth difference beam and an elevation difference beam from the echoes received from the blood vessel of interest;
- ii) modulate the directions of the transmitted and received sonic energy based upon the sum, azimuth difference and elevation difference beams in order to lock on to the highest Doppler energy calculated from echoes from the flow of blood in the blood vessel of interest, and
- 25 iii) calculate the three dimensional position of the highest Doppler energy from the blood flow in the vessel of interest.

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Optionally, the processor can also be programmed to determine at least one additional beam having an angle between the azimuth difference beam and the elevation difference beam prior to modulating the directions of the transmitted and received sonic energy, wherein the at least one additional beam is used to modulate the directions of the transmitted and received sonic energy. Naturally, the angle of the at least one additional beam can vary. In a particular embodiment, the at least one additional beam is at an angle that is orthogonal to the blood vessel of interest.

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Moreover, the present invention extends to a method as described above, wherein steps (b) through (e) are periodically repeated so that the three dimensional position of blood flow in the vessel of interest is tracked, and the parameter of blood flow is periodically calculated and displayed on the display monitor. In a particular
5 embodiment, the period of time between repeating steps (b) through (e) is sufficiently short so that the parameter being measured remains constant, e.g., 20 milliseconds.

The present invention further extends to a method for determining a parameter of blood flow in a particular region of a blood vessel of interest, comprising the steps of:

- 10 a) providing an array of sonic transducer elements, wherein the element spacing in the array is greater than, equal or less than a half wavelength of the sonic energy produced by the elements, wherein at least one element transmits sonic energy, and a portion of the elements receive sonic energy;
- 15 b) directing sonic energy produced by the at least one element of the array into a volume of the subject's body having the particular region of the blood vessel of interest,
- 20 c) receiving echoes of the sonic energy from the volume of the subject's body having the particular region of the blood vessel of interest;
- d) reporting the echoes to a processor programmed to
 - 25 i) Doppler process the echoes to determine radial velocity of the blood flowing in the particular region of the blood vessel of interest;
 - ii) calculate a three dimensional position of blood flow in the particular region of the blood vessel of interest; and
 - iii) calculate the parameter of blood flow in the particular region of the
30 blood vessel of interest at the three dimensional position calculated in step (ii); and
- (e) displaying the parameter on a display monitor that is electrically connected to the processor.

A particular method of calculating the three dimensional position of blood flow in such a method of the present invention comprises having the processor programmed to:

- i) determine a sum beam, an azimuth difference beam and an elevation difference beam from the echoes received from the
5- _____ particular region of the blood vessel of interest;
- ii) modulate the directions of the transmitted and received sonic energy based upon the sum, azimuth difference and elevation difference beams in order to lock on to the highest Doppler energy calculated from echoes received
10 _____ -from the flow of blood in the particular region of the blood vessel of interest, and
- iii) calculate the three dimensional position of the highest Doppler energy from the blood flow in the particular region of the blood vessel of
15 _____ interest.

As explained above, at least one additional beam can also be determined and used to calculate the three dimensional position.

20 Furthermore, the present invention extends to a method for determining a parameter of blood flow in a blood vessel of interest, comprising the steps of:

- a) providing an array of sonic transducer elements, wherein the element spacing in the array is greater than, equal or less than a half wavelength of the sonic energy
25 _____ produced by the elements, wherein at least one element transmits sonic energy, and a portion of elements receive sonic energy;
- b) directing sonic energy produced by the at least one element of the array into a volume of the subject's body having the blood vessel of interest,
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- c) receiving echoes of the sonic energy from the volume of the subject's body having the blood vessel of interest;
- d) reporting the echoes to a processor electrically connected to the elements of
35 _____ the array, wherein the processor is programmed to

- i) Doppler process the echoes to determine radial velocity of the blood flowing in the blood vessel of interest;
- 5 ii) determine a sum beam, an azimuth difference beam and an elevation difference beam from the echoes received from the blood vessel of interest;
- 10 iii) modulate the directions of the transmitted and received sonic energy based upon the sum, azimuth difference and elevation difference beams in order to lock on to the highest Doppler energy calculated from echoes from the flow of blood in the blood vessel of interest,
- 15 iv) calculate the three dimensional position of the highest Doppler energy from the blood flow in the vessel of interest; and
- v) calculate the parameter of blood flow in the blood vessel at the three dimensional position calculated in step (iv); and
- (e) displaying the parameter on a display monitor that is electrically connected to the processor.
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As explained above, an operator performing a method of the present invention can obtain blood flow parameters from a blood vessel of interest, and even from a particular region of a blood vessel of interest.

25 Moreover, the present invention extends to a method for determining a parameter of blood flow in a particular region of a blood vessel of interest, comprising the steps of:

- a) providing an array of sonic transducer elements, wherein the element spacing in the array is greater than, equal or less than a half wavelength of the sonic energy produced by the elements, wherein at least one element transmits sonic energy, and a portion of the elements receive sonic energy;
- 30 b) directing sonic energy produced by the at least one element of the array into 35 a volume of the subject's body having the particular region of the blood vessel of interest,

- c) receiving echoes of the sonic energy from the volume of the subject's body having the particular region of blood vessel of interest;
- 5 d) reporting the echoes to a processor electrically connected to the elements of the array, wherein the processor is programmed to
- 10 i) Doppler process the echoes to determine radial velocity of the blood flowing in the particular region of the blood vessel of interest;
- 15 ii) determine a sum beam, an azimuth difference beam and an elevation difference beam from the echoes received from the particular region of the blood vessel of interest;
- 20 iii) modulate the directions of the transmitted and received sonic energy based upon the sum, azimuth difference and elevation difference beams in order to lock on to the highest Doppler energy calculated from echoes from the flow of blood in the particular region of the blood vessel of interest,
- 25 iv) calculate the three dimensional position of the highest Doppler energy from the blood flow in the particular region of the blood vessel of interest; and
- 30 v) calculate the parameter of blood flow in the particular region of the blood vessel at the three dimensional position calculated in step (iv); and
- (e) displaying the parameter on a display monitor that is electrically connected to the processor.

In another embodiment, the present invention extends to a device for determining a parameter of blood flow in a blood vessel of interest, comprising:

- 35 a) an array of sonic transducer elements, wherein the element spacing in the array is greater than, equal or less than a half wavelength of the sonic energy

produced by the elements, and at least one element transmits sonic energy, and a portion of the elements receive sonic energy;

- 5 b) a processor electrically connected to the array so that echoes received from a volume of the subject's body having the blood vessel of interest due to directing sonic energy produced by the at least one element of the array into the subject's body is reported to the processor, wherein the processor is programmed to:
- 10 i) Doppler process the echoes to determine radial velocity of the blood flowing in the blood vessel of interest;
 - ii) calculate a three dimensional position of blood flow in the blood vessel of interest; and
 - iii) calculate the parameter of blood flow in the blood vessel of interest at the three dimensional position calculated in step (ii); and
- 15 (c) a display monitor that is electrically connected to the processor which displays the parameter of blood flow calculated by the processor.

A parameter of blood that can be determined with a device of the present invention includes blood flow volume, vector velocity, Doppler spectral distribution, etc. The
20 parameter being measured can be an instantaneous value, or an average value determined over a heart cycle.

Moreover, the present invention extends to a device as described above, wherein the processor is programmed to:

- 25 i) determine a sum beam, an azimuth difference beam and an elevation difference beam from the echoes received from the blood vessel of interest after Doppler processing the echoes;
- 30 ii) modulate the directions of the transmitted and received sonic energy based upon the sum, azimuth difference and elevation difference beams in order to lock on to the highest Doppler energy calculated from echoes from the flow of blood in the blood vessel of interest,
- 35 iii) calculate the three dimensional position of the highest Doppler energy from the blood flow in the vessel of interest; and

- iv) calculate the parameter of blood flow in the blood vessel of interest at the three dimensional position calculated in (iii).

Optionally, a processor of a device of the present invention can be further
5 programmed to determine at least one additional beam having an angle between the azimuth difference beam and the elevation difference beam prior to modulating the directions of the transmitted and received sonic energy, wherein the at least one additional beam is used to modulate the directions of the transmitted and received sonic energy. In a particular embodiment, the at least one additional beam is at an
10 angle that is orthogonal to the blood vessel of interest.

Moreover, in a another embodiment of a device of the present invention, the distance between the elements of the array is greater than $1/2$ the wavelength of the sonic energy generated by the at least one element

15 Furthermore, the present invention extends to a device for determining a parameter of blood flow in a blood vessel of interest, comprising:

a) an array of sonic transducer elements, wherein the element spacing in the
20 array is greater than, equal or less than a half wavelength of the sonic energy produced by the elements, and at least one element transmits sonic energy, and portion of the elements receive sonic energy;

b) a processor electrically connected to the array so that echoes received from a
25 volume of the subject's body having the blood vessel of interest due to directing sonic energy produced by the at least one element of the array into the subject's body is reported to the processor, wherein the processor is programmed to:

- i) Doppler process the echoes to determine radial velocity of the blood flowing in the blood vessel of interest;
- 30 ii) calculate a three dimensional position of blood flow in the blood vessel of interest; and
- iii) calculate the parameter of blood flow in the blood vessel of interest at the three dimensional position calculated in step (ii)

35 (c) a display monitor that is electrically connected to the processor which displays the parameter of blood flow calculated by the processor.

Particular parameters of blood flow that can be determined with a device of the present invention include, but certainly are not limited to blood flow volume, vector velocity, and Doppler spectral distribution. The parameter being measured can be an
5 instantaneous value, or an average value determined over a heart cycle.

In addition, a processor of a device of the present invention can be further programmed to determine at least one additional beam having an angle between the azimuth difference beam and the elevation difference beam prior to modulating the
10 directions of the transmitted and received sonic energy, wherein the at least one additional beam is used to modulate the directions of the transmitted and received sonic energy. In a particular embodiment, the at least one additional beam is at an angle that is orthogonal to the blood vessel of interest.

15 Moreover, the present invention extends to a method for generating a three dimensional image using sonic energy of a blood vessel of interest in a subject, the method comprising the steps of:

- a) providing an array of sonic transducer elements, wherein the element spacing
20 in the array is greater than, equal or less than a half wavelength of the sonic energy produced by the elements, wherein at least one element transmits sonic energy, and a portion of the elements receive sonic energy;
- b) directing sonic energy produced by the at least one element of the array into
25 a volume of the subject's body having the blood vessel of interest,
- c) receiving echoes of the sonic energy from the volume of the subject's body having the blood vessel of interest;
- d) reporting the echoes to a processor programmed to
 - i) Doppler process the echoes to determine radial velocity of the blood flowing
in the blood vessel of interest;
 - ii) calculate a three dimensional position of blood flow in the blood vessel
35 of interest;

- iii) repeat steps (i) through (ii) to generate a plurality of calculated three dimensional positions; and
- vi) generate a three dimensional image of the blood vessel of interest from the plurality of calculated three dimensional positions; and

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(e) displaying the three dimensional image on a display monitor that is electrically connected to the processor.

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Furthermore, the present invention permits an operator utilizing a method of the present invention to generate a three dimensional image of not only a blood vessel in the body, but even a particular region of a blood vessel in the body.

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Numerous means available for calculating the three dimensional position of a blood vessel and even a particular portion of a blood vessel are encompassed by the present invention. A particular means comprises having the programmed processor:

- i) determine a sum beam, an azimuth difference beam and an elevation difference beam from the echoes received from the blood vessel of interest after Doppler processing the echoes;
- ii) modulate the directions of the transmitted and received sonic energy based upon the sum, azimuth difference and elevation difference beams in order to lock on to the highest Doppler energy calculated from echoes from the flow of blood in the blood vessel of interest, and
- iii) calculate the three dimensional position of the highest Doppler energy from the blood flow in the vessel of interest, and
- iv) repeat steps (i) through (iii) to generate a plurality of calculated three dimensional positions.

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Optionally, a processor of a method of the present invention can also be programmed to determine at least one additional beam having an angle between the azimuth difference beam and the elevation difference beam prior to modulating the directions of the transmitted and received sonic energy, and the at least one additional beam is also used to modulate the directions of the transmitted and

received sonic energy, and calculate the three dimensional position of the highest Doppler energy. In a particular embodiment, the at least one additional beam is at an angle that is orthogonal to the blood vessel of interest.

5 The present invention also extends to a method for generating a three dimensional image of a blood vessel of interest in a subject using sonic energy, the method comprising the steps of:

- 10 a) providing an array of sonic transducer elements, wherein the element spacing in the array is greater than, equal or less than a half wavelength of the sonic energy produced by the elements, wherein at least one element transmits sonic energy, and a portion of the elements receive sonic energy;
- b) directing sonic energy produced by the at least one element of the array into
15 a volume of the subject's body having the blood vessel of interest,
- c) receiving echoes of the sonic energy from the volume of the subject's body having the blood vessel of interest;
- 20 d) reporting the echoes to a processor programmed to
 - i) Doppler process the echoes to determine radial velocity of the blood flowing in the blood vessel of interest;
 - 25 ii) determine a sum beam, an azimuth difference beam and an elevation difference beam from the echoes received from a portion of the blood vessel of interest;
 - iii) modulate the directions of the transmitted and received sonic energy
30 based upon the sum, azimuth difference and elevation difference beams in order to lock on to the highest Doppler energy calculated from echoes from the flow of blood in the blood vessel of interest,
 - iv) calculate the three dimensional position of the highest Doppler energy
35 from the blood flow in the vessel of interest; and

- v) repeat steps (i) through (iv) to generate a plurality of calculated three dimensional positions;
 - vi) generate a three dimensional image of the blood vessel of interest from
5 _____ the plurality of calculated three dimensional positions; and
- (e) displaying the three dimensional image on a display monitor that is electrically connected to the processor.
- 10 Optionally, the three dimensional image can be of a particular region of a blood vessel of interest. Moreover, a processor of a method described herein can also determine at least one additional beam having an angle between the azimuth difference beam and the elevation difference beam prior to modulating the directions of the transmitted and received sonic energy, and the at least one additional beam is
- 15 also used to modulate the directions of the transmitted and received sonic energy, and calculate the three dimensional position of the highest Doppler energy. Angles for use with the at least one additional beam are described above.
- Moreover, in another embodiment of the present invention, the distance between the
- 20 elements of the array is greater than $1/2$ the wavelength of the sonic energy generated by the at least one element.
- Furthermore, the present invention extends to a device generating a three dimensional image of a blood vessel of interest in a subject using sonic energy,
- 25 comprising:
- a) an array of sonic transducer elements, wherein the element spacing in the array is greater than, equal or less than a half wavelength of the sonic energy produced by the elements, and at least one element transmits sonic energy, and a
 - 30 portion of the elements receive sonic energy;
 - b) a processor electrically connected to the array so that echoes received from a volume of the subject's body having the blood vessel of interest due to directing sonic energy produced by the at least one element of the array into the subject's body is
 - 35 reported to the processor, wherein the processor is programmed to:

- i) Doppler process the echoes to determine radial velocity of the blood flowing in the blood vessel of interest;
- ii) calculate a three dimensional position of blood flow in the blood vessel of interest;
- 5 iii) repeat steps (i) through (ii) to generate a plurality of calculated three dimensional positions;
- v) generate a three dimensional image from the plurality of calculated three dimensional positions, and
- 10 (c) a display monitor that is electrically connected to the processor which displays the three dimensional image.

As explained above, a device of the present invention permits an operator to generate and display three dimensional images of a blood vessel of interest, and
 15 even of a particular region of a blood vessel that the operator wants to investigate closely. Moreover, in a particular embodiment, a processor of a device of the present invention can be programmed to calculate the three dimensional position of a blood vessel by

- 20 i) determining a sum beam, an azimuth difference beam and an elevation difference beam from the echoes received from the blood vessel of interest after Doppler processing the echoes;
- ii) modulating the directions of the transmitted and received sonic energy
 25 based upon the sum, azimuth difference and elevation difference beams in order to lock on to the highest Doppler energy calculated from echoes from the flow of blood in the blood vessel of interest,
- iii) calculating the three dimensional position of the highest Doppler
 30 energy from the blood flow in the vessel of interest; and
- iv) repeat steps (I) through (iii) in order to generate a plurality of calculated three dimensional positions used to generate the three dimensional image.

Optionally, the processor can be programmed to further determine at least one additional beam having an angle between the azimuth difference beam and the elevation difference beam prior to modulating the directions of the transmitted and received sonic energy, wherein the at least one additional beam is used to modulate

5 the directions of the transmitted and received sonic energy. The angle between the azimuth difference beam and the elevation difference beam of the additional beam can vary. In a particular embodiment, the at least one additional beam is at an angle that is orthogonal to the blood vessel of interest.

10 Furthermore, the present invention extends to a thinned array for use in an ultrasound device, comprising a plurality of sonic transducer elements, wherein the element spacing in the array is greater than a half wavelength of the sonic energy produced by the elements, and the elements are positioned and sized within the array, and sonic energy is electronically steered by the elements so that any grating

15 lobes produced by the sonic energy are suppressed. In a particular embodiment, the elements positioned and sized so that they are flush against each other.

Hence, the current invention performs blood velocity monitoring by collecting Doppler data in three dimensions; azimuth, elevation, and range (depth); so that the point (in

20 three dimensional space) at which the velocity is to be monitored can be acquired and tracked when the patient or the sensor moves. The invention also produces a three dimensional map of the blood flow and converts measured radial velocity to true vector velocity.

25 Moreover, in this invention, once the desired signal is found, it will be precisely located and continually tracked with accuracy far better than the resolution. A heretofore unknown method to achieve sub-resolution tracking and mapping involves a novel and unobvious extension of a procedure called "monopulse". Monopulse tracking has been used in military applications for precisely locating and tracking a

30 point target with electromagnetic radiation. However, it has never been utilized in connection with sonic waves to determine the velocity of moving fluids *in vivo*.

This invention provides: (1) affordable three-dimensional imaging of blood flow using a low-profile easily-attached transducer pad, (2) real-time vector velocity, and (3)

35 long-term unattended Doppler-ultrasound monitoring in spite of motion of the patient

or pad. None of these three features are possible with current ultrasound equipment or technology.

The pad and associated processor collects and Doppler processes ultrasound blood velocity data in a three-dimensional region through the use of a two-dimensional phased array of piezoelectric elements on a planar, cylindrical, or spherical surface. Through use of unique beamforming and tracking techniques, the invention locks onto and tracks the points in three-dimensional space that produce the locally maximum blood velocity signals. The integrated coordinates of points acquired by the accurate tracking process is used to form a three-dimensional map of blood vessels and provide a display that can be used to select multiple points of interest for expanded data collection and for long term continuous and unattended blood flow monitoring. The three dimensional map allows for the calculation of vector velocity from measured radial Doppler.

In a particular embodiment, a thinned array (greater than half-wavelength element spacing of the transducer array) is used to make a device of the present invention inexpensive and allow the pad to have a low profile (fewer connecting cables for a given spatial resolution). The array is thinned without reducing the receiver area by limiting the angular field of view. The special 2-0 phased array used in this invention makes blood velocity monitoring inexpensive and practical by (1) forming the beams needed for tracking and for re-acquiring the blood velocity signal and by (2) allowing for an element placement that is significantly coarser than normal half-wavelength element spacing. The limited range of angles that the array must search allows for much less than the normal half wavelength spacing without reducing the total receiver area.

Grating lobes due to array thinning can be reduced by using wide bandwidth and time delay steering. The array, or at least one element of the array, is used to sequentially insonate the beam positions. Once the region of interest has been imaged and coarsely mapped, the array is focused at a particular location on a particular blood vessel for measurement and tracking. Selection of the point or points to be measured and tracked can be based on information obtained via mapping and may be user guided or fully automatic. Selection can be based, for example, on peak response within a range of Doppler frequencies at or near an approximate location.

In the tracking mode a few receiver beams are formed at a time: sum, azimuth difference, elevation difference, and perhaps, additional difference beams, at angles other than azimuth (= 0 degrees) and elevation (= 90 degrees). Monopulse is applied at angles other than 0 and 90 degrees (for example 0, 45, 90, and 135

5 degrees) in order to locate a vessel in a direction perpendicular to the vessel. When the desired (i.e. peak) blood velocity signal is not in the output, this is instantly recognized (e.g., a monopulse ratio, formed after Doppler filtering, becomes non-zero) and the array is used to track (slow movement) or re-acquire (fast movement) the desired signal. Re-acquisition is achieved by returning to step one to form and

10 Doppler-process a plurality of beams in order to select the beam (and the time delay or "range gate") with the most high-Doppler (high blood velocity) energy. This is followed by post-Doppler monopulse tracking to lock a beam and range gate on to the exact location of the peak velocity signal. In applications such as transcranial Doppler, where angular resolution based on wavelength and aperture size is

15 inadequate, fine mapping is achieved, for example, by post-Doppler monopulse tracking each range cell of each vessel, and recording the coordinates and monopulse-pair angle describing the location and orientation of the monopulse null. With a three-dimensional map available, true vector velocity can be computed. For accurate vector flow measurement, the monopulse difference is computed in a

20 direction orthogonal to the vessel by digitally rotating until a line in the azimuth-elevation or C-scan display is parallel to the vessel being monitored. The aperture is more easily rotated in software (as opposed to physically rotating the transducer array) if the aperture is approximately circular (or elliptical) rather than square (or rectangular). Also, lower sidelobes result by removing elements from the four

25 corners of a square or rectangular array in order to make the array an octagon.

In this invention, as long as (1) a blood vessel or (2) a flow region of a given velocity can be resolved by finding a 3-D resolution cell through which only a single vessel passes, that vessel or flow component can then be very accurately located within the cell. Monopulse is merely an example of one way to attain such sub-resolution

30 accuracy (SRA). Other methods involve "super-resolution" or "parametric" techniques used in "modern spectral estimation", including the MUSIC algorithm and autoregressive modeling, for example. SRA allows an extremely accurate map of 3 - D flow.

Furthermore, the present invention utilizes post-Doppler, sub-resolution tracking and

35 mapping; it does Doppler processing first and uses only high Doppler-frequency data.

This results in extended targets since the active vessels approximate “lines” as opposed to “points”. In three-dimensional space, these vessels are resolved, one from another, At a particular range, the monopulse angle axis can be rotated (in the azimuth-elevation plane) so that the “line” becomes a “point” in the monopulse angle

5 direction. That point can then be located by using super-resolution techniques or by using a simple technique such as monopulse. By making many such measurements an accurate 3-D map of the blood vessels results.

Methods for extending the angular field of view of the thinned array (that is limited by grating lobes) include (1) using multiple panels of transducers with multiplexed
10 processing channels, (2) convex V-shaped transducer panels, (3) cylindrical shaped transducer panel, (4) spherical shaped transducer panel, and (5) negative ultrasound lens. If needed, moving the probe and correlating the sub-images can create a map of an even larger region.

Active digital ~~beamforming~~ beamforming can also be utilized, but the implementation depends on a
15 choice to be made between wideband and narrowband implementations. If emphasis is on high resolution mapping of the blood vessels, then a wide bandwidth (e.g., 50% of the nominal frequency) is used for fine range resolution. If emphasis is on Doppler spectral analysis, measurement, and monitoring, the map is only a tool. In this case, a narrowband, low cost, low range-resolution, high sensitivity
20 implementation might be preferred. A wideband implementation would benefit in performance (higher resolution, wider field of view, and reduced grating lobes) using time-delay steering while a narrowband implementation would benefit in cost using phase-shift steering. The invention can thus be described in terms of two preferred implementations.

25 In a wideband implementation, time delay steering can be implemented digitally for both transmit and receive by over-sampling and digitally delaying in discrete sample intervals. In a narrowband implementation, (1) phase steering can be implemented digitally (digital beamforming) for both transmit and receive, and (2) bandpass sampling (sampling at a rate lower than the signal frequency) can be employed with
30 digital down-conversion and filtering.

In traditional ultrasound applications the transmit signal is a short series of constant amplitude pulses. The pulses are 50 percent duty cycle pulse of acoustic energy at a single frequency. The amplitude must be very consistent or system performance will suffer. More advanced ultrasound techniques require shaping the transmitted beam by

35 applying different power levels to different elements of an array of transmitter
elements. However, the need to change voltage from one series of pulses to the next
can cause problems with the electronics associated with the transmission. One problem
is there is not enough time for the voltage to settle from one series to the next. This has
the effect of causing artifacts in the image.

One means of overcoming the problems is to hold the voltage constant and modify a
series of pulses to achieve the different power levels. Two techniques can be used with
the pulses to control the power. The first is using fewer pulses in the series and the
second is modifying the width of the pulses, thus, pulse width modulation. When using
5 fewer pulses, the start of the series is delayed to position the center of series in the
correct relationship with the other series. This invention employs both techniques. By
varying both number and width a wide range of power levels can be achieved. The
bandwidth characteristic of the piezoelectric element and system smoothes the pulse
width modulated signal into a lower amplitude signal with greatly reduced harmonics.
10 A second benefit is the reduced complexity of the transmit circuitry. Without this
invention a significant amount of high power low impedance circuitry is required to
rapidly change and hold the voltage at a new level. To do this with precision adds even
more circuitry. The amount of digital logic necessary to create this new pulse width
modulated transmit signal is small and does not significantly affect the cost of the
15 system.

Accordingly, it is an object of the present invention to locate the point in three
dimensional space having the greatest high-Doppler energy, and determining
coordinates for that point. With that information, and the radial velocity of the blood
2035 flowing through the blood vessel at that point, a variety of blood flow parameters can

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be calculated at that point, including, but not limited to vector velocity of blood flow,
volume of blood flow, or Doppler spectral distribution. The parameter being measured can
be an instantaneous value, or an average value determined over a heart cycle.

5
25 It is also an object of the present invention to continuously track and map *in vivo* the
point in three dimensional space having the greatest Doppler-energy, and using the
coordinates to generate a three dimensional image of a blood vessel and blood flow
therein that possess a much greater resolution than images generated using
40 heretofore known Doppler ultrasound methods and devices.
30

It is yet another object of the present invention to provide a thinned array which does not utilize the number of element transducers as are required with heretofore known Doppler ultrasound devices. As a result, the decreased number of elements in the
15 array decreases size of the array utilized and provides a patient being analyzed with
35 mobility that would not be available if using conventional ultrasound devices to obtain blood flow parameters such as vector velocity, blood flow volume, and Doppler spectral distribution. The parameter being measured can be an instantaneous value, or an average value determined over a heart cycle.

20 These and other aspects of the present invention will be better appreciated by reference
5 to the following drawings and Detailed Description.

BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 illustrates the Blood Flow Mapping Monitor in use with a Transcranial Doppler
1025 Probe, as an example.

FIG.2 shows a 64-element bistatic ultrasound transducer array example, where, with $D = 2d$, the same elements are reconfigured differently for transmit and receive during the acquisition phase of operation. Fig. 2 (a) shows the Receive Configuration, where all 64
15 elements receive at once. Fig. 2 (b) shows the Transmit
30 Configuration, where, during acquisition, the 16 sub-apertures transmit one at a time.

FIG. 3 is an example overall block diagram of a blood flow mapping monitor embodiment.

- 20 FIG. 4 illustrates ultrasound beam coverage for the TCD array example of Fig. 2. The left illustration shows 25 digitally beam-formed beams, as an example. On the right, is shown, for that example, the manner in which the transmit beam encompasses 21 receive beams in the acquisition mode.
- 25 FIG. 5 shows one-dimensional patterns for a bistatic transducer array with $D = 2d$ as in FIG. 2. FIG. 5a (top) shows the transmit element pattern. FIG. 5b shows the receive Element Pattern and Array Pattern with the receiver beam steered to broadside ($x=0$). The Array Pattern has Grating Lobes (Receiver Ambiguities). FIG. 5c shows the resultant two-way beam pattern (product of all three patterns above).
- 4030 The Grating Lobes are suppressed.
- FIG. 6 is the same as FIG. 5, with the receive array beam steered to $x = 0.2$.
- FIG. 7 shows the Two-way pattern of a receiver beam steered to the half power point ($x=0.2$). This is Fig. 6c plotted in dB.
- 35 FIG.8 shows a one-dimensional representation of the example of FIG. 4. FIG. 8a shows the product of transmit and receive Element Patterns. FIG. 8b plots a set of five receive beams showing Grating Lobes of the Thinned Array. FIG. 8c plots the resultant two-way beams with Grating Lobes suppressed.
- FIG.9 is a block diagram of one possible embodiment of the Transmit-Receive Electronics for a Bistatic Ultrasound Imaging Sensor and Blood Monitoring Monitor.
- 205 FIG. 10 shows the receiver channel signal spectrum illustrating functions performed by the FPGA of Fig. 9 on each of the 64 received signals for a narrowband case.
- FIG.11 shows the geometry involved in using azimuth monopulse to more accurately determine the cross-range location of a vessel. The range resolution is better than the cross-range resolution and the measured radial velocity field or color flow map
- 10 has been utilized to rotate and orient the azimuth and elevation axes so that the center of the vessel is vertical, at approximately zero azimuth. The black circular cylinder represents the location of all points within the spatial resolution cell that have a particular velocity.
- 25
- 15 FIG. 12 shows the geometry involved in using Doppler ultrasound to determine the diameter of a vessel or the velocity field within the vessel. While the initial 3-D orientation of the vessel is general, a measured 3-D radial velocity field or 3-D color flow map has been utilized to rotate and orient the azimuth and elevation axes so

that the center of the vessel is vertical, at approximately zero azimuth. In other words, the coordinate system has been rotated about the depth-axis so that the centerline of the vessel is in the depth-elevation plane. This can be accomplished either by a change of coordinates in software or by physically rotating the ultrasound

5 probe. The black circular cylinder represents the location of all points within the
25 -illustrated box that have a particular velocity. The diameter of the cylinder is then measured as the azimuth extent of a high-resolution depth-azimuth or B-scan image at the Doppler frequency under examination.

FIG. 13 illustrates the Blood Flow Mapping Monitor in use with a Transcranial 10 Doppler Probe, as an example.

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FIG. 14 shows a 52-element ultrasound transducer array example, based on an 8 by 8 rectangular array of elements with 3 elements removed from each corner to make the array octagonal instead of rectangular or square. For this example, the elements are square ($d_1 = d_2 = d$) and $L/d = 8$.

35+5 FIG. 15 shows a typical pattern of electronically scanned beams produced by the array in FIG. 14. The beam width is nominally, given by the signal wavelength divided by the size, L , of the array. The angular field of view (F.O.V.) is limited by the maximum angle to which the array can be steered without producing grating lobes that are not sufficiently attenuated by the pattern of the individual $d \times d$ element.

20_5 FIG. 16 shows one-dimensional patterns for an eight-element monostatic linear transducer array corresponding to a column or a row in FIG. 16. FIG. 16a (top) shows the Element Pattern and Array Pattern with the beam steered to broadside ($x=0$). The Array Pattern has Grating Lobes (Receiver Ambiguities). FIG. 16b shows the resultant beam pattern. The Grating Lobes are suppressed.

25_10 FIG. 17 is the same as FIG. 16, with the array beam steered to an angle at which a grating lobe exceeds the highest sidelobe. The thinned array of Figure 16 should not be steered beyond $\pm \arcsin(\lambda/2L)$ ($\pm 4.70^\circ$ for the example used) if grating lobes are to be suppressed.

FIG. 18 shows the pattern of a beam steered to the point where the grating lobe
30-15 problem appears. This is Fig. 17b plotted in dB.

FIG. 19 shows a dual 52-active-element ultrasound transducer array example (similar to that in FIG. 14) with a total of 116 elements, 52 of which are used at a time. Figure 19 B shows that the two sub-arrays are in two different planes, tilted to reduce

the overlap between beams from the two sub-arrays and maximize the azimuth angular field of view.

FIG. 20 shows a 52-active-element ultrasound transducer array example (similar to that in FIG. 14) with a total of 84 elements (52 of which are used at a time) and with a slightly
25 convex cylindrical shape. The indicated $L_1 \times L_2$ sub-aperture would be activated for the formation of beams pointed to one side.

FIG. 21 is an example overall block diagram of a blood flow mapping monitor embodiment.

FIG. 22 is a block diagram of one possible embodiment of the analog Transmit-Receive
40-30 Electronics for an Ultrasound Imaging Sensor and Blood Monitor.

FIG. 23 shows the geometry involved in using azimuth monopulse to more accurately determine the cross-range location of a vessel. The measured radial velocity field or color flow map has been utilized to rotate and orient the azimuth and elevation axes so that the center of the vessel is vertical, at approximately zero azimuth. The black circular cylinder
1535 represents the location of all points within the spatial resolution cell that have a particular velocity.

Figure 24A illustrates a series of control signals for transmitting a 50 percent duty cycle burst of ultrasonic energy at a reference amplitude.

Figure 24B illustrates a series of control signals that implement the prior art method of reducing the burst of energy by reducing the amplitude of the control signal.

Figure 24C illustrates a series of control signals that implement the herein disclosed method of reducing the burst energy by reducing the width of the pluses to significantly less than 50 percent duty cycle.

5 Figure 24D illustrates a series of control signals that product a reduction of bursts of energy by also reducing the number of pulses in the burst, while keeping the center of the burst in correct temporal relationship with other bursts.

10 Figure 25 is a circuit diagram of an exemplary apparatus for controlling and transmitting a series of bursts of varying amplitude.

Figure 26 is a circuit diagram of an exemplary transmission apparatus for varying burst of power by controlling pulse width and/or number of pluses per burst.

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DETAILED DESCRIPTION OF THE INVENTION

The invention involves (1) a family of ultrasound sensors, (2) the interplay of a set of core technologies that are unique by themselves, and (3) a number of design options which
20 represent different ways to implement the invention. To facilitate an organizational understanding of this many-faceted invention, a discussion of each of the three topics above follows.

The sensors addressed are all two-dimensional (i.e., planar or on the surface of a convex shape such as a section of a cylinder) arrays of piezoelectric crystals for use in active,
25 non-invasive, instantaneous (or real-time), three-dimensional imaging and monitoring of blood flow. The sensors use a unique approach to 3-D imaging of blood velocity and blood flow that (1) allows for finer image resolution than would otherwise be possible with the same hardware complexity (number of input cables and associated electronics) and (2) allows for finer accuracy than would ordinarily be possible based on the
30 resolution. The invention measures and monitors 3-D vector velocity rather than merely the radial component of velocity.

Moreover, the present invention also utilizes (1) array thinning with large elements and limited scanning, (2) array shapes to reduce peak sidelobes and extend the field of coverage, (3) post-Doppler sub-resolution tracking, (4) post-Doppler sub-resolution mapping

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mapping, (5)
35 additional methods for maximizing the angular field of view, and (6) various digital beamforming procedures for implementing the mapping, tracking, and measurement processes. The present invention also extends to array thinning, where the separation between array elements is significantly larger than half the wavelength.
5 This reduces the number of input cables and input signals to be processed while maintaining high resolution and sensitivity and avoiding ambiguities. In a transcranial Doppler application, for example, where signal to noise and hence receiver array area is of paramount importance, array thinning is possible without reducing the receiver array area because a relatively small (compared to other applications) angular field of view is
10 needed.

Thinning with full aperture area imposes limitations on the angular field of view. Methods for expanding the field of view include using more elements than are active at any one time. For example, if the electronics are switched between two identical panels, the cross-range field of view at any depth is increased by the size of the panel. If the
15 panels are pointed in slightly different directions so that overlapping or redundant beams are

avoided, the field of view is doubled. A generalization of this approach involves the use of an array on a cylindrical or spherical surface.

20 Once a section of a blood vessel is resolved from other vessels in Doppler, depth, and two angles (az and el), Post-Doppler sub-resolution processing locates that section to an accuracy that is one-tenth to one-twentieth of the resolution. This allows for precise tracking and accurate mapping. Tracking provides for the possibility of unattended long term monitoring and mapping aids the operator in selecting the point or points to be monitored.

25 Furthermore, methods of the present invention permit non-invasive, continuous, unattended, volumetric, blood vessel tracking, ultrasound monitoring and diagnostic device for blood flow. It will enable unattended and continuous blood velocity measurement and monitoring as well as 3-dimensional vascular tracking and mapping using an easily attached, electronically steered, transducer probe that can be in the form of a small pad for monitoring application, when desired. Moreover, a device and method of the present
30 invention have applications in measuring the parameters described above in any part of the body. A nonlimiting example described below involves a cranial application. However as set forth, a device and method of the present have applications in any part of the body, and can be used to track and map any blood vessel in the body. A device of the present invention can, for example:

1. Measure and continuously monitor blood velocity with a small low-profile probe that can be adhered, lightly taped, strapped, banded, or otherwise easily attached to the portion of the body where the vascular diagnosis or monitoring is required.
- 5 2. Track and maintain focus on multiple desired blood vessels in spite of movement.
3. Map 3-D blood flow; e.g., in the Circle of Willis (the central network of arteries that feeds the brain) or other critical vessels in the cranial volume.
4. Perform color velocity imaging and display a 3-D image of blood flow that is
10 rotated via track ball or joystick until a desired view is selected.
- 10 5. Form and display a choice of projection, slice, or perspective views, including (1) a projection on a depth-azimuth plane, a B-scan, or a downward-looking perspective, (2) a projection on an azimuth-elevation plane, a C-scan, or a forward-looking perspective, or (3) a projection on an arbitrary plane, an
15 arbitrary slice, or an arbitrary perspective.
6. Use a track ball and buttons to position circle markers on the points where
measurement or monitoring of vector velocity is desired.
7. Move the track location along the blood vessel by using the track ball to slide the circle marker along the image of the vessel.
- 20 8. Display actual instantaneous and/or average vector velocity and/or estimated average volume flow.
9. Maintain a multi-day history and display average blood velocity versus time for each monitored vessel over many hours.
10. Sound an alarm when maximum or minimum velocity is exceeded or when
25 emboli count is high; and maintain a log of emboli detected.
11. Track, map, and monitor small vessels (e.g., 1mm in diameter), resolve vessels as close as 4 mm apart (for example), and locate them with an accuracy of ± 0.1 mm, for example.
- 30 Moreover, as explained herein, numerous methods have applications in obtaining the three dimensional coordinates of points along a blood vessel from echoes returned from the body, and are encompassed by the present invention. A particular nonlimiting example of such a method having applications herein is a novel and unobvious variation of monopulse tracking. For tracking purposes utilizing

monopulse, up to nine beams are simultaneously formed for each transmit beam position. In addition to the “sum” beam that corresponds to the transmitted beam, there will either be 4 monopulse difference beams or there will be 8 overlapping focused beams. If a cluster of eight focused beams is used, these will be highly

5 overlapped with the sum beam, and displaced very slightly from the sum beam, with
their centers equally spaced on a small circle around the center of the sum beam. These
satellite beams would then operate in pairs to form four difference beams. For example,
the azimuth Monopulse ratio can be produced in two different ways, which will call
“liner” and “non-linear”. The non-linear method will determine the
10 magnitudes or the powers of three received signals, left, right, and sum (L , R , and S),
and compute $M_3 ([L] - [R]) / [S]$. The linear method uses complex signals and
computes the azimuth monopulse ratio as the real part of the ratio D_a/S , where $D_a = L$
- R . D_a is the azimuth difference.

15 For an ideal point target, the linear method for computing M_a results in an excellent
estimate of the azimuth angle error. It also has the advantage of only requiring 4, instead
of 8 auxiliary beams. These 4 beams would be an azimuth difference beam, D_a , an
elevation difference beam, and two diagonal difference beams. The individual beams,
such as L and R , are not needed. However, beam shapes will be highly
20 distorted by refraction through bone and tissue, and a “sub-optimum” non-linear
approach might be more robust.

Regardless of which monopulse method is used, the conventional two difference
beams used in radar (azimuth difference and elevation difference) may not be
25 enough. The projection of the high-velocity data on a plane perpendicular to the
transducer line of sight (the C-scan) will usually be a line, not a point. With multiple
difference beams, equally spaced in angle, one will be approximately perpendicular to
the C-scan projection of the vessel. The system will select the monopulse difference
output with the largest magnitude. This provides an approximate
30 orientation of the C-scan projection of the vessel. The corresponding monopulse ratio
(provided the sum beam power exceeds a threshold) is used to correctly resteer and
maintain a beam precisely centered on that vessel.

If the power map output of a Wall filter is used for the monopulse beams, the beam
35 outputs are power and hence a complex ratio is not available. In that case the nonlinear
method would be used. An alternative is to use the complex wall filter

output, before computing the power, with the linear method. During measurement, however, the output of a particular (high velocity) FFT Doppler bin may be used for monopulse (provided that the magnitude or power of the sum beam at that Doppler exceeds a threshold). In that case either the linear or the nonlinear monopulse ratio may be used.

Another alternative is to use FFT processing and form the monopulse ratio (linearly or non-linearly) at the output of a high-velocity Doppler-frequency cell with high sum-beam power. For example, set a power threshold and select the highest (positive or negative) velocity cell with power that exceeds the threshold. Since the data in a single FFT cell is expected to be noisy, this procedure is recommended for a measurement dwell, where enough time is spent in a single beam position to have both useable velocity resolution and the ability to make several measurements (multiple FFT's per frame).

FFT-Based Monopulse and Monopulse Averaging During Measurement

In a K pulse dwell, let $K=K_1 \times K_2$ where K_1 is the number of input pulses used in the FFT and K_2 is the number of FFTs. Instead of performing monopulse to re-steer the beam every K_1 pulses, we compute the monopulse ratio at the output of a desired high velocity Doppler bin, and average its value over K_2 FFTs. This reduces the steering noise while assuring that we are locating the center of the vessel (the highest Doppler Energy). We chose the highest Doppler frequency for which the minimum sum beam power exceeds a threshold, and utilize only that Doppler cell for monopulse. The average is best performed as a weighted average. For example, if D_n and S_n are (say, elevation) difference-beam and sum beam outputs in the n th FFT for the selected Doppler bin, we chose:

$$M = \frac{\sum_{n=1}^{K_2} |S_n|^2 M_n}{\sum_{n=1}^{K_2} |S_n|^2}, \text{ where } M_n = \text{Re} \{D_n / S_n\}$$

$$\text{or } M_n = \frac{|D_n|^2}{|S_n|^2}$$

depending on whether linear or non-linear monopulse is used. For linear monopulse it might be best to use only one large FFT ($K_2=1$). For non-linear monopulse, the expression simplifies to:

$$M = \frac{\sum_{n=1}^{K_2} |D_n|^2}{\sum_{n=1}^{K_2} |S_n|^2}$$

[Note that because a ratio is involved (so that beam pointing error is not confused with signal strength) even the “linear” method is non-linear.]

A device of the present invention will allow a person with little training to apply the sensor and position it based on an easily understood ultrasound image display. The unique sensor can continuously monitor artery blood velocity and volume flow for early detection of critical events. It will have an extremely low profile for easy attachment, and can track selected vessels; e.g., the middle cerebral artery (MCA), with no moving parts. If the sensor is pointed to the general volume location of the desired blood vessel (e.g., within ± 1 cm.), it will lock to within ± 0.1 mm of the point of maximum radial component of blood flow and remain locked in spite of patient movement.

A device of the present invention can remain focused on the selected blood vessels regardless of patient movement because it produces and digitally analyzes, in real time, a 5-dimensional data base composed of signal-return amplitude as a function of:

1. Depth,
2. Azimuth,
3. Elevation,
4. Radial component of blood velocity,
5. Time.

Since a device of the present invention can automatically locate and lock onto the point with the maximum volume of blood having a significant radial velocity, unattended continuous blood velocity monitoring is one of its uses. By using the precise relative location of the point at which lock occurs as a function of depth, a device of the present invention can map the network of blood vessels as a 3-

dimensional track without the hardware and computational complexity required to form a conventional ultrasound image. Using the radial component of velocity along with the three-dimensional blood path, a device of the present invention can directly compute vector velocity.

5 A device used in a method of the present invention is a non-mechanical Doppler ultrasound-imaging sensor comprising probes, processing electronics, and display. Specific choices of probes allow the system to be used for transcranial Doppler (TCD), cardiac, dialysis, and other applications.

10 The present invention may be better understood by reference to the following non-limiting Examples, which are provided as exemplary of the invention. The following Examples are presented in order to more fully illustrate particular embodiments of the invention. They should in no way be construed, however, as limiting the broad scope of the invention.

15 **EXAMPLE 1 AN ULTRASOUND DIAGNOSTIC AND MONITORING SENSOR WITH REAL-TIME 3-D MAPPING AND TRACKING OF BLOOD FLOW**

20 This embodiment of the present invention has application for medical evaluation and monitoring multiple locations in the body; however, the transcranial Doppler application will be used as an example to describe the invention.

This invention provides: (1) affordable three-dimensional imaging of blood flow using a low-profile easily-attached transducer pad, (2) real-time vector velocity, and (3) long-term unattended Doppler-ultrasound monitoring in spite
25 of motion of the patient or pad. None of these three features are possible with current ultrasound equipment or technology.

The pad and associated processor collects and Doppler processes ultrasound blood velocity data in a three dimensional region through the use of a planar phased array of piezoelectric elements. Through use of unique beamforming
30 and tracking techniques, the invention locks onto and tracks the points in three-dimensional space that produce the locally maximum blood velocity signals. The integrated coordinates of points acquired by the accurate tracking process is used to form a three-dimensional map of blood vessels

and provide a display that can be used to select multiple points of interest for expanded data collection and for long term continuous and unattended blood flow monitoring. The three dimensional map allows for the calculation of vector velocity from measured radial Doppler.

- 5 A thinned array (greater than half-wavelength element spacing of the transducer array) is used to make the device inexpensive and allow the pad to have a low profile (fewer connecting cables for a given spatial resolution). The same physical array can also be used to form a broad transmit beam encompassing a plurality of narrow receive beams. Initial acquisition of the blood velocity signal is attained
- 10 ~~blood velocity signal is attained~~ by insonating a large region by defocusing the transmit array or by using a small transmitting sub-aperture, for example. The computer simultaneously applies numerous sets of delays and/or complex weights to the receiver elements in order to form M simultaneous beams. With M beams being formed simultaneously, the receiver can dwell M times as long, so as to obtain high S/N and fine Doppler
- 15 ~~as long, so as to obtain high S/N and fine Doppler~~ resolution. For an embodiment that utilizes a small transmitting sub-aperture, the source of the transmitted energy within the array (i.e., the location of the transmitter sub-aperture) varies with time in order to lower the temporal average spatial peak intensity to prevent skin heating.
- 20 The array is thinned without reducing the receiver area by limiting the angular field of view. When needed, a map of a larger region is created by moving the probe and correlating the sub-images. Once the region of interest has been imaged and coarsely mapped, the full transmitter array is focused at a particular location on a particular blood vessel for tracking. In the tracking
- 25 mode: (1) grating lobes due to array thinning are reduced by using wide bandwidth and time delay steering and (2) only three beams are formed at a time: sum, azimuth difference, and elevation difference. When the desired (i.e. peak) blood velocity signal is not in the output, this is instantly recognized (e.g., a monopulse ratio, formed after Doppler filtering, becomes non-zero)
- 30 and the array is used to track (slow movement) or re-acquire (fast movement) the desired signal. Re-acquisition is achieved by returning to step one to form and Doppler-process a plurality of beams in order to select the beam (and the

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time delay or “range gate”) with the most high-Doppler (high blood velocity) energy. This is followed by post-Doppler monopulse tracking in azimuth, elevation, and range to lock a beam and range gate on to the exact location of the peak velocity signal.

5 In applications such as transcranial Doppler, where angular resolution based on wavelength and aperture size is inadequate, fine mapping is achieved, for example, by post-Doppler monopulse tracking each range cell of each vessel, and recording the coordinates describing the location of the monopulse null. With a three-dimensional map available, true vector velocity can be

10 computed. For accurate vector flow measurement, the monopulse difference is computed in a direction orthogonal to the vessel by digitally rotating until a line in the azimuth-elevation or C-scan display is parallel to the vessel being monitored.

All current ultrasound devices (including “Doppler color flow mapping”

15 systems) form images that are limited by their resolution. In some applications, such as TCD, the low frequency required for penetration makes the azimuth and elevation resolution at the depths of interest larger than the vessel diameter. In this invention, as long as (1) a blood vessel or (2) a flow region of a given velocity can be resolved by finding a 3-D resolution cell

20 through which only a single vessel passes, that vessel or flow component can then be very accurately located within the cell. Monopulse is merely an example of one way to attain such sub-resolution accuracy (SRA). SRA allows an extremely accurate map of 3-D flow.

This invention utilizes post-Doppler, sub-resolution tracking and mapping; it

25 does Doppler processing first and uses only high Doppler-frequency data. This results in extended targets since the active vessels approximate “lines” as opposed to “points”. In three-dimensional space, these vessels are resolved, one from another. At a particular range, the azimuth-elevation axis can be rotated so that the “line” becomes a “point” in the azimuth dimension.

30 That point can then be located by using super-resolution techniques or by using a simple technique such as monopulse.

Overview of the Embodiment

The invention is complex because it involves (1) a family of ultrasound sensors (for different parts of the body), (2) the interplay of a set of core technologies that are unique by themselves, and (3) a number of design

5 options which represent different ways to implement the invention. To facilitate an organizational understanding of this many-faceted invention, we precede a description of an overall preferred embodiment with a discussion of each of the three topics above.

The sensors addressed are all two-dimensional (i.e., planar) arrays of
10 piezoelectric crystals for use in active, non-invasive, instantaneous (or realtime), three-dimensional imaging and monitoring of blood flow. While the sensors and the techniques for their use apply to all blood vessels in the body, the figures and detailed description emphasizes the transcranial Doppler (TCD) monitor because that application is most difficult to implement
15 without all of the components of this invention. The sensors use a unique approach to 3-D imaging of blood velocity and blood flow that (1) allows for finer image resolution than would otherwise be possible with the same hardware complexity (number of input cables and associated electronics) and (2) allows for finer accuracy than would ordinarily be possible based on the
20 resolution. The invention measures and monitors 3-D vector velocity rather than merely the radial component of velocity.

The core technologies that constitute the invention are (1) array thinning with suppression of ambiguities or grating lobes, (2) post-Doppler sub-resolution tracking, and (3) post-Doppler sub-resolution mapping. The invention
25 encompasses two ways to thin the array (reducing the number of input cables and input signals to be processed while maintaining high resolution and avoiding ambiguities). The first is bistatic operation: the second is broadband operation. In the TCD application, where signal to noise and hence receiver array area is of paramount importance, array thinning is possible without
30 reducing the receiver array area because a relatively small (compared to other applications) angular field of view is needed. One particular bistatic approach to thinning reduces transmitter area and consequently poses a

problem of excessive spatial peak intensity (skin heating) in the TCD application. This is solved by a component invention called transmitter diversity (which lowers the temporal average of the spatial peak intensity). The phase-defocusing bistatic approach and the monostatic or bistatic

5 broadband approach to thinning all use the entire aperture and hence do not require transmitter diversity.

In the TCD application, the achievable angular resolution is poor, regardless of the method of thinning, or whether or not thinning is used. Once a section of a blood vessel is resolved from other vessels in Doppler, depth, and two
10 angles (az and el), Post-Doppler sub-resolution processing locates that section to an accuracy that is 10 to 20 times as fine as the resolution. This allows for precise tracking and accurate mapping. Tracking provides for the possibility of unattended long term monitoring and mapping aids the operator in selecting the point or points to be monitored.

15 There are many options available in the design of any member of the family of sensors that utilizes any or all of the core technologies that comprise this invention. A two-dimensional array is established art that can be designed in many ways and can have many sizes and shapes (rectangular, round, etc.). Digital beamforming (DBF) is a technique that has been in the engineering
20 literature (especially radar and sonar) for many years. One medical ultrasound DBF patent cites many references, while another describes a particular instance of DBF without citing the other patent or any other prior art. While planar arrays, DBF, Doppler ultrasound, and color flow imaging are prior art, the manner in this specification of using such established
25 technologies to map, track, measure, and monitor blood flow is unique.

The embodiment is a non-invasive, continuous, unattended, volumetric, blood vessel tracking, ultrasound monitoring and diagnostic device. It will enable unattended and continuous blood velocity measurement and monitoring as well as 3-dimensional
530 ~~well as 3-dimensional~~ vascular tracking and mapping using an easily attached, electronically steered, transducer probe that can be in the form of a small pad for monitoring application, when desired. Although the device has application to multiple body parts, the cranial application will be used as a specific example. The device can, for example:

40-35 1. Measure and continuously monitor blood velocity with a small low-profile probe that can be adhered, lightly taped, strapped, banded, or otherwise easily attached to the portion of the body where the vascular diagnosis or monitoring is required.

2. Track and maintain focus on up to four desired blood vessels in spite of
+5 movement.
3. Map 3-D blood flow; e.g., in the Circle of Willis (the central network of
arteries that feeds the brain).
4. Perform color velocity imaging and display a 3-D image of blood flow that
is rotated via track ball or joystick until a desired view is selected.
- 210 5. Form and display a choice of projection, slice, or perspective views,
including (1) a projection on a depth-azimuth plane, a B-scan, or a downward-
looking perspective, (2) a projection on an azimuth-elevation plane, a C-scan, or a
forward-looking perspective, or (3) a projection on an arbitrary plane, an arbitrary
slice, or an arbitrary perspective.
- 215 6. Use a track ball and buttons to position circle markers on the points at which
we wish to measure and monitor vector velocity.
7. Move the spatial resolution cell being measured along the blood vessel by
using the track ball to slide the circle marker along the image of the vessel.
8. Display actual instantaneous and/or average vector velocity and/or
320 estimated average volume flow.

9. Maintain a 3-day history and display average blood velocity versus time for each monitored vessel over 14 hours.

10. Sound an alarm when maximum or minimum velocity is exceeded or when emboli count is high.

25 11. Track, map, and monitor vessels as small as 1mm in diameter, resolve vessels as close as 4 mm apart (for example), and locate them with an accuracy of ± 0.1 mm.

The Monitoring Device will allow a person with little training to apply the sensor and position it based on an easily understood ultrasound image

430 display. The unique sensor can continuously monitor artery blood velocity and volume flow for early detection of critical events. It will have an extremely low profile for easy attachment, and can track selected vessels; e.g., the middle cerebral artery (MCA), with no moving parts. If the sensor is pointed to the general volume location of the desired artery (e.g., within ± 0.5 cm.), it will lock

435 to within ± 0.1 mm of the point of maximum radial blood flow and remain locked in spite of patient movement.

The device can remain focused on the selected blood vessels regardless of patient movement because it produces and digitally analyzes, in real time, a 5-dimensional data base composed of signal-return amplitude as a function

205 of:

6. Depth, 2. Azimuth, 3. Elevation, 4. Radial blood velocity, 5. Time. Since the device can automatically locate and lock onto the point with the maximum volume of blood having a significant radial velocity, unattended continuous blood velocity monitoring is one of its uses. By using the precise

2510 relative location of the point at which lock occurs as a function of depth, the device can map the network of blood vessels as a 3-dimensional track without the hardware and computational complexity required to form a conventional ultrasound image. Using radial velocity along with the three-dimensional blood path, the device can directly compute vector velocity.

3015 The proposed device is a non-mechanical Doppler ultrasound-imaging sensor consisting of probes, processing electronics, and display. Specific choices of

probes allow the system to be used for transcranial Doppler (TCD), cardiac, dialysis, and other applications.

Fig. I shows the TCD configuration and the initial definition of the display
20 -screen. The TCD system is comprised of one or two probes attached to the
5 head with a "telephone operator's band" or a Velcro strap. The interface and
processing electronics is contained within a small sized computer. A thin cable
containing 64 micro coax cables attaches the probe to the electronics in the
computer. When the operator positions the probe on the head the Anterior,
25 -Middle and Posterior Cerebral Arteries and the Circle of Willis are
40 imaged on the screen along with other blood vessels. The arteries or vessels of
interest are selected by viewing the image. The system locks onto the blood
vessels and tracks their position electronically. A variety of selected parameters
is presented on the screen; e.g., the velocity, the pulse rate, depth of region
30 -imaged, gain and power level. Using only one probe the TCD can
45 monitor up to two arteries (vessels) at a time. Presented on the screen are dual
traces, one for each artery. The blood velocity can be dynamically monitored.
As shown in Fig. I both the current blood velocity (dark traces) and any
historic trace (lighter color) can be displayed simultaneously. The average
35 -blood velocity or estimated average flow for each artery is displayed
20 below the respective velocity trace. The image shows the arteries and the channel
used for each artery. When two probes are used, the display is split showing
signals from both of them. Using a different probe (i.e., different size) with the
same electronics and display, the unit can be used to measure and monitor the
blood flow in a carotid artery. Similarly, it can be used to perform
25 this function for dialysis, anesthesia, and in other procedures.
5 The sensor is a two dimensional array of transducer elements (piezoelectric
crystals) that are configured and utilized differently for transmit and receive
during acquisition. For example, if a square ($N \times N$) array is used, all N^2
elements would receive at the same time, but only a 2×2 sub-aperture would
30 transmit at any one time. This is illustrated in FIG. 2 for the case of $N=8$. The
10 -array need not be square. Any $M \times N$ array may be utilized in this manner. All
 NM received signals (64 in our example) are sampled, digitized, and

processed. This can be done, for example, in a desk top or lap top personal computer with additional cards for electronics and real-time signal processing as illustrated in FIG. 1 and FIG. 3. If the PCI bus in FIG 2 becomes a bottleneck for
 15 -high speed processing, a pipelined or systolic architecture
 5 would be used. Alternatively, the processing can be performed in an application specific integrated circuit (ASIC).

The small (4 element) transmit sub-aperture (FIG. 2 b) produces a broad transmit beam that insonates a region containing many receive beams. This is
 20 -schematically illustrated in FIG. 4 for the particular case of a square array
 40 and square elements such as in FIG. 2. Since data is received from each element of the array, this data can be combined in a processor (FIG. 3, for example) in many different ways to form any number of beams. The transmitter is larger than a single array element so that it can provide some
 25 -selectivity and *not* insonate the grating lobes caused by array thinning
 45 (spacing the array elements more than $\lambda/2$ wavelength apart). The concept is illustrated below for a 1-dimensional array forming a beam that measures only one angle. For a two-dimensional array, this represents a horizontal or vertical cut through the cluster of beams shown in FIG. 4. FIG. 4 was an approximate
 30 and conceptual representation of the two-angle (azimuth and
 20 elevation) extension of the single angle case detailed below.

“Grating lobes” are ambiguities or extra, unwanted, beams caused by using a transducer array whose elements are too large and hence too far apart. The following analysis illustrates grating lobe suppression for the worst case of
 35 -narrowband signals and phase-shift beam processing. Time delay processing
 25 using wideband signals would be similar, but would further attenuate or eliminate grating lobes, resulting in even better performance.

The next four figures show beam pattern amplitudes plotted against

$$x(d/\lambda) \sin \theta, \quad (1)$$

30 5 where x represents a normalization for the angle, θ , from which reflected acoustic energy arrives. The azimuth (or elevation) angle, θ , is zero in the

broadside direction, perpendicular to the transducer array. The width (or length) of a transmitter is $2d$, where d is the width (or length) of a single element of the
 10 -receiver array. The wavelength of the radiated acoustic wave is $\lambda = c/f$, where c is the acoustic propagation velocity (1540 meters/second in
 soft tissue) and f is the acoustic frequency (usually between 1 and 10 megahertz). FIG. 5a shows the transmitter pattern

$$15 \quad a_T(x) = \sin 2\pi x / 2\pi x \quad (2)$$

40 for the special case of uniform insonation over the $2d$ -wide transmitter sub-aperture being used.

The receiver pattern is the product of the receiver element pattern and the receiver array pattern

$$20 \quad a_R(x) = a_{RE}(x) a_{RA}(x) \quad (3)$$

Each of these two component patterns is plotted separately in FIG. 5b. Again assuming the special case of a uniform receiver element (and a square element in the case of a 2-D array), the receive element pattern is

$$205 \quad a_{RE}(x) = \sin \pi x / \pi x. \quad (4)$$

The receiver element pattern is twice as wide as the transmitter pattern because the receiver element is half as wide as the transmitter. In the far-
 2530 field, i.e., for $\lambda r \gg L^2$, where r is the range or depth and L is the length of the aperture, the receive array pattern steered to the angle $\theta = \theta_0$ is

$$a_{RA}(x) = \sum_{n=0}^{N-1} w_n e^{j2\pi n(x-x_0)}, \quad (5)$$

305 where w_n is a weighting to reduce sidelobes and N is the number of elements in one dimension. As seen in Figure 5b, equation (5) is periodic in x . The peak at $x=x_0$ ($x_0 = 0$ in Fig 5) is the desired beam and the others are grating lobes.

In the near field, when focused at (r_0, θ_0) , equation (5) is replaced by the slightly better general Fresnel approximation:

$$a_{RA}(x, z) = \sum_{n=0}^{N-1} w_n e^{j2\pi \left[n(x-x_0) + \left(n - \frac{N-1}{2} \right)^2 (z-z_0) \right]} \quad (6)$$

5

(provided that the range significantly exceeds the array size, $r \gg L$), where $x = d \sin \theta / \lambda$, as before, and

$$z = d^2 \cos^2 \theta / \lambda r. \quad (7)$$

10

Because the receiver aperture is sampled with a spatial period of d , the receiver array pattern will be periodic in $\sin \theta$, with a period of λ/d (equation 5). This periodicity means that the array pattern is ambiguous. When the array is pointed broadside ($\theta = 0$), it will also be pointed at the angle $\theta = \sin^{-1}(\lambda/d)$, for example. In terms of the normalized variable, x , the period is unity. Since $|\sin \theta|$ cannot exceed 1, the variable x is confined to the interval $[-d/\lambda, d/\lambda]$.

15

The conventional element spacing is $d = \lambda/2$. Thus, in a conventional phased array, x is always between -0.5 and +0.5, and hence ambiguities are not encountered. In a highly thinned array ($d > \lambda$), there will normally be ambiguities or grating lobes as illustrated in FIG. 5b. The second grating lobe, at $x=2$ or $\theta = \sin^{-1}(2\lambda/d)$, is not real when d does not exceed 2λ .

20

FIG. 5c shows the two-way pattern. The grating lobe suppression, resulting from the choice of a transmitter diameter of $D = 2d$ is valid for all values of d . In a two dimensional array, the elements could be rectangular instead of square ($d_x \times d_y$), and the results would still be valid. Similar results could be obtained for an array in which the elements are staggered from row to row (and/or column to column). For example, if the receiver array is a "bathroom tile" of hexagonal elements, the transmitters could be chosen as sub-arrays consisting of an element and its six surrounding neighbors.

25

In FIG. 6 the same array is used as in FIG. 5, but the receiver element signals are combined with a phase taper that steers the beam to $x = 0.2$. This is approximately (a little less than) the half power point, where $a_I(x) a_{re}(x) =$

0.707. In FIG. 6c, we see that the grating lobes are not completely suppressed, with the largest one at $x = -1 + 2 = -.8$. FIG. 7 shows this in decibels. The worst-case grating lobe is attenuated by at least 25 dB, even in the stressing case of extremely narrow band operation. A Hanning window

5 was applied to keep the sidelobes lower than the peak grating lobe. These Figures were produced in MATLAB, using the following software (rn-file):

```

x= -2:1/64:2 -1/64;
p=pi*x+eps; R=sin(p) ./p;
10 p=2*p; T=sin(p) ./p;
N=8
n=0:N-1;
% xo=0;
xo=0.2; % is 2-way 1/2 power
15 e=exp (j*n'*2*pi*(x-xo));
w=hanning(N);
% E=(1/N)* ones(1, N)*e;
E= (2/N) *w'*e;
subplot(311); plot (x,abs (T));
20 subplot(312); plot (x, [abs (R) ;abs (E)]); I
TRE=abs (T) . *abs (R) . abs (E) ;
subplot(313); plot(x,TRE);
figure(2); plot(x,20*log10 (TRE));
zoom on;

```

25 The dimensions in FIG 4 are representative for a transcranial Doppler application of the invention, to provide a specific example. If $f = 2$ MHz is chosen for the center frequency, the wavelength is 0.77 mm. An 8x8 array with a width and/or length of $L=1$ cm, provides a one dimensional thinning

30 ratio of $2 d/\lambda = 3.247$. For a square array, the total number of elements is reduced by a factor of $(2 d/\lambda)^2 \geq 10$ from that of a filled array. Even greater thinning ratios are possible. Even if d/λ is kept less than 2 to avoid a second grating lobe (at $x=2$), complexity reductions up to a factor of 16 are possible. For the 1 cm array at 2 MHz, the hyperfocal distance (where the 3 dB focal region extends to infinity)

region extends to infinity) is $L^2/4\lambda = 3.25$ cm. Thus, a fixed focus probe suffices for this application. However, since the simultaneous formation of multiple receive beams is conveniently performed digitally, dynamic focus on receive is easily accomplished.

_____ The quadratic phase distribution across the elements required to focus
55 ~~elements required to focus~~ in depth is simply added to the linear phase
distributions required to steer
-the beams.

FIG. 8a shows the product of the transmitter pattern (FIG. 5a or 6a) and the
receiver element pattern. FIG. 8b plots the element patterns for a set of five
10 beams steered to $x = -.2, -.1, 0, .1, \text{ and } .2$. This set of five receive beams
shows grating lobes of the thinned array. FIG. 8c shows the set of resulting 2-
way patterns obtained by multiplying the patterns in FIG. 8b by the function
plotted in FIG. 8a. Here, the grating lobes are suppressed. This represents a
horizontal or vertical cut through the cluster of beams in FIG. 4.

15 Using the configuration described above, the cluster of beams in Figs. 4 and
8c is used to approximately locate the desired point for collecting the blood
velocity signal. For example the output of each beam in the cluster would be
Doppler processed by performing an FFT or equivalent transformation on a
20 sequence of pulse returns. The pulse repetition frequency (PRF) would
typically be less than or equal to 9 kHz to unambiguously achieve a depth of 8.5
cm for the TCD application. In order to obtain a velocity resolution as fine as L_w
 $= 1$ cm per second (to distinguish brain death), a dwell of duration $T = \lambda / (2\Delta v)$
 $= 38.5$ ms, or 347 pulses at 9kHz, is desired. For efficient FFT
25 processing, the number of pulses used would be zero filled to a power of 2 such as
512.

The example shown in Figs. 2 through 8 was an 8 by 8 receiver array forming a 5
by 5 cluster of beams. This is an example of an approximate rule of
30 thumb for this invention, that an N element linear array is recommended for use
in producing $N/2 + 1$ beams for N even and $[N+1]/2$ beams for N odd. Thus, a 16
by 10 element rectangular array would preferably be used to form a 9 by 6
cluster of beams, though the actual number of beams formed is arbitrary. This
recommended number of beams is derived below.

If an N elements were used to form orthogonal beams, e.g., by an N -point FFT, then there would be N beams in a 180° angular region, from -90° to $+90^\circ$, corresponding to $-1 < u < 1$, where $u = \sin \theta$. In conventional phased array
 5 ultrasound, a 128 ($=N$) element array is used to produce 256 ($=2N$) lines (sequentially scanned beams) in a 90° angular region from -45° to 45° , corresponding to $-.707 < u < .707$. If the array is filled, then $x = u/2$ (Equation 1) and $2N$ beams are conventionally formed in $|x| < \sqrt{2}/4$. When we thin the array, we prefer to have $|x| < 0.2 = 1/5$ (the 3 dB point of the curve in Fig. 7a). The
 10 number of beams in that region, for the same beam density as used in current practice, is given by

$$\text{Recommended No. of beams} = (1/5)N \div (\sqrt{2}/4) = 2\sqrt{2} N/5 \approx 0.5657 N.$$

15 The beams are formed digitally, using software on a personal computer or using digital signal processing hardware to implement equations such as Equation 5 or 6. The electronic interface between the probe and the processor is diagrammed in Fig. 9. This figure illustrates the case of signals from 64 elements being connected to a single A/D converter, and power being
 20 applied to sets of four elements. The use of a separate ND converter for every received channel, for example, is another possible implementation of this invention.

A conventional, half-wavelength spaced, monostatic, phased array could
 25 sequentially search a region of interest, but it would require far more elements and would thus be far more costly. Using the array differently in transmit and receive, not only allows for the formation of multiple beams; it also enables the use of the angular pattern of the transmitter to suppress receiver grating lobes. This allows for a "thinned" array (elements spaced less than a half
 30 wavelength apart). Because receive beams are formed only in a limited angular region, a wide-angle receiver element pattern (which usually implies a small element) is not required. In fact, the size of the receiver element can be as large as the element spacing. Thus the receiver array is "thinned" only in the sense that the element spacing exceeds a half wavelength. Since the

element size also exceeds a half wavelength, the array area is not reduced. It is thinned only in terms of number of elements, not in terms of receiver area. Consequently, there is no reduction in signal-to-noise ratio, nor a requirement for increased transmitter power.

5

A monostatic array would transmit from the full aperture, scanning the transmitted beam over the region being examined. The “bistatic” array of this invention transmits from a sub-aperture to insonate multiple receive beam positions simultaneously. Since there is an FDA limit to spatial peak, temporal average, intensity (I_{spta}), there may be a danger of exceeding this limit at the transducer surface, creating a danger of burning the skin. This potential danger is eliminated by using a different transmit sub-aperture for each coherent dwell or burst of pulses. This transmitter diversity technique spreads the temporal average intensity over the face of the array, reducing I_{spta} to what it would be if the entire array were used at once.

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For the particular implementation pictured in FIG. 9, an ND converter is multiplexed amongst the 64 elements. The signal spectrum at any of these elements is centered at $f_0 = 2$ MHz, as shown in FIG. 10a. This is a real signal with a spectrum that is symmetric about $f = 0$. This analog signal is bandpass filtered (BPF) to insure that there is little power outside of a 444 kHz band centered at 2 MHz. If a $512/9 = 56.889$ MHz ND converter is used, each receive channel is sampled at $f_s = 888.9$ kHz, giving rise to a real sampled signal with a spectrum as shown in FIG. 10b. A processing element such as a field programmable gate array (FPGA) is used to shift the frequency by $f_s/4$ (FIG. 10c) by “multiplying” by quarter cycle samples of sinusoids (which are zeros and ones). The same FPGA also digitally filters (or Hubert transforms) the complex signal to decimate its sampling rate by a factor of two. The spectrum of the decimated digital low-pass signal is shown in FIG. 10 d.

The signal sent to the processor from each element has the spectrum shown in Fig. 9d, and consists of $r = f_s/2$ complex samples per second. The total data rate into the processor is approximately 57 megabytes per second. For non-real-time operation, tens of seconds of data at a time will be collected in

system memory and then transferred to hard disk. For real-time monopulse tracking, only three beams are formed, so that the data rate is reduced to $3 \times 0.8889 = 2.67$ Mbytes, or 5.33 Mbytes allowing for bit growth.

- 5 The transmitted pulses are sent to a group of four elements. The particular embodiment shown in Figure 9 uses diodes to block the received signals and prevent mutual coupling between the four receive elements. After a coherent pulse train (or pulse burst used for Doppler processing), the waveform is switched to another set of 4 elements for the next burst. A separate power
10 amplifier is associated with each of the 16 sets of elements so that the switching can be accomplished at low power.

- One embodiment of sub-resolution tracking (i.e., tracking and locating blood flow to a small fraction of a spatial resolution cell) is "Monopulse". Monopulse tracking is performed as follows. A particular set of complex weights are
15 applied to the set of received signals (64 in the example of FIG 2) to steer a beam at the middle cerebral artery, for example. The phase taper across the array defines the steering direction and the amplitude taper (called a window in radar and a shading in sonar) is used to provide low sidelobes for high dynamic range. The beam output (a linear combination of the signals) is
20 range gated (time delay corresponding to the desired depth) and the range-gated / beam-formed output from a sequence of transmitted pulses is then Fourier transformed to obtain a plot of amplitude versus Doppler frequency. The receive beam is steered digitally to the point that produces the maximum amplitude at high Doppler frequencies.

- 25 Since the measured data at each element is stored, the digital processor can apply more than one set of weights at a time, forming more than one beam. For software monopulse the processor will form three beams, all in the same direction. All three beams may have the same phases applied to the element signals; but the amplitudes will differ. The beam called *Sum* has all positive
30 amplitudes, with the larger weights applied to the central elements. This forms a fairly broad beam. The beam called *Az* for "azimuth difference beam" has large positive weights on the rightmost elements and large negative weights on the leftmost elements (or vice versa). The beam called *El* for

elevation difference has large positive weights on the top-most elements and large negative weights on the bottom-most elements. A correctly pointed beam would have $Az = El = 0$, and Sum would be maximized.

The ratio of the peak Doppler amplitude outputs: Az/Sum , is a precise measure of the azimuth pointing error and the corresponding ratio El/Sum measures the elevation pointing error. The digital steering phase taper is thus corrected with data from a single burst of pulses. The duration of the pulse burst is the reciprocal of the medically required Doppler resolution (usually corresponding to the minimum blood velocity that can support life). Without techniques such as those described in this specification, a sequence of at least four additional Doppler dwells or pulse bursts would be required (above, below, to the right, and to the left) in a hunt and seek method to find the correct (maximum peak Doppler Amplitude) beam. With monopulse, the correction is very precise (to within ± 0.1 mm of the point of maximum peak Doppler amplitude) and virtually instantaneous. For the bistatic digitally beamformed sensor, the original data exists in computer memory. Hence, whenever the Doppler processed monopulse differences are non zero, the same data set could even be reprocessed to form a correctly pointed beam. A slower processor would merely process the **next** burst correctly.

A “front view” perspective display or a C scan display (azimuth horizontal and elevation vertical) of the blood flow map at the desired range will allow someone to aim the transducer probe or pad at the desired point (highest amplitude for high Doppler), so that the desired point is initially within the center beam. The receiver array is then steered electronically so that the monopulse differences are zero and hence the central beam is precisely aimed at the desired point. Slight motions are corrected using monopulse and large motions are corrected by again forming all beams to re-acquire the peak signal. All corrections are made entirely electronically, in the data processing or digital beamforming. A narrow receiver beam will always be precisely pointed (to within a tenth or 20^{th} of the receiver beamwidth) as long as the desired point remains within the much larger region covered by the transmitter (FIG. 4).

35 True vector velocity is computed from the blood vessel map and the radial
 velocity measured from the pulse Doppler dwell. A map, far more accurate than
 that attainable with the available angular resolution is attained as follows. The
 low-resolution map is used to locate a vessel of interest and a beam is
 5 locked on it at a fixed range, using azimuth and elevation monopulse. The
 coordinates of the point at which lock occurs is recorded. The range is then changed
 slightly, another lock (on the same vessel) is obtained, and the coordinates are
 recorded. In this manner, the vessel is mapped far more accurately than would be
 predicted from the available image resolution. All
 10 vessels within the field of view of the probe are similarly mapped. By moving the
 probe angle slightly, another region can be mapped in the same manner. Several
 such maps can be correlated over the region of pair-wise overlap and converted to a
 common coordinate system. In this manner a larger region is mapped and displayed
 than that of the current field of view. The current field
 15 of view would be highlighted, outlined, or presented as a color flow map. Points
 to be monitored in the current field are then selected by moving a cursor along
 the display (point and click). The selected points are Doppler processed and
 tracked using three-dimensional monopulse. While Doppler measurements
 provide only the radial component of velocity, the accurate
 20 blood vessel map provides the exact three-dimensional orientation of the vessel at
 the point being monitored. The measured radial velocity is divided by the
 projection of a unit vector representing the vessel at the monitored point onto the
 transducer line of sight. This gives the magnitude of the true vector blood
 velocity.
 25 Sub-resolution mapping accuracy is attainable if (1) the range-azimuth-elevation-
 Doppler resolution cell being examined encompasses only a single blood vessel,
 and (2) "azimuth" monopulse is performed with the usually vertical e-axis tilted
 so that the orientation of the vessel in the spatial resolution cell being processed
 is parallel to the e-r plane ("azimuth" is
 30 constant).

The user will ascertain from the display, that the resolution cell being
 monitored contains only a single vessel, and would rotate the 3-0 blood-

vessel map to a C-scan aspect (elevation up and azimuth to the right). A vertical mark will appear in the display, within the resolution circle, to signify the orientation of the monopulse axis. This axis (parallel to the line separating the positively and negatively weighted array elements) can then be oriented

5 so that the mark is aligned with the blood vessel in either of two ways. The probe
can be physically twisted (rotated about the line of sight), or it can be electronically
rotated via digital processing because the weights are applied digitally.

FIG. 11 illustrates the segment of a vessel in a single resolution cell, after
10 rotation. The resolution cell shown is not a cube because the range resolution
will typically be finer than the cross-range resolution. The illustrated circular
cylinder represents blood cells in a vessel reflecting energy at a fixed Doppler
frequency. These represent a cylindrical annulus of blood cells, at a constant
distance from the vessel wall, moving with approximately the same velocity.
15 In the single resolution cell of Fig. 11, the return at the highest Doppler would
represent a line in three-dimensional space (the axis of the vessel) and hence a
point on the azimuth axis after rotation. When applied to the highest Doppler
output, the Sum beam would have broad peak at zero azimuth ($a=0$) and the
monopulse ratio, $r=Az/Sum$, will be a linear function of the azimuth
20 angle to which the array is phase steered:

$$r(a) = ka.$$

This result can be attained by applying the same phase across the aperture for the
Az and *Sum* beams, but using the derivative of the *Sum* beam amplitude weights
with respect to x and y respectively for the *Az* and *El* aperture weights.

25 **Other Embodiments**

If the wide transmit beam (for search and acquisition) is created by using a
quadratic phase curvature instead the scheme of Fig. 2b, transmitter diversity
may not be needed. Furthermore the manner of controlling grating lobes in Fig.1
and Figs. 5 - 8 is only one of many. Using a wider bandwidth and time-
30 delay steering can also reduce grating lobes.

EXAMPLE II ULTRASOUND MEASUREMENT OF BLOOD VOLUME FLOW

As explained above, current ultrasound Doppler devices measure radial velocity. Several methods now exist for 3-D ultrasound imaging, such as those involving

5 transducer motion. A three-dimensional image with Doppler allows for the measurement of vector velocity. Example I above provides measurement and long term monitoring of three-dimensional vector velocity. If the resolution of a color flow Doppler image is sufficient to provide an estimate of the inside diameter of the blood vessel, then measurement of volume blood flow becomes practical. Presently

10 available ultrasound imaging devices have either low resolution or they only produce a two-dimensional image. The present invention combines vector velocity information (such as attained as explained in Example I above) with additional information to obtain volume flow. The additional information is the inside diameter of the vessel under examination, the blood velocity profile across the vessel, or the

15 vector velocity as a function of time and position (i.e., the velocity field). This additional information can be obtained from a high-resolution radial-Doppler or color flow image or from external data such as a high-resolution MRI image.

A two-dimensional array of piezoelectric elements, or some other means, is used to image blood flow in a three dimensional region. A particular point on a particular

20 vessel is selected and the vector representing the orientation of the vessel is noted. The radial velocity divided by the cosine of the angle made by the vessel with the line of sight at the measurement point is the magnitude of the vector velocity. That number integrated over the vessel cross section would give the volume flow in volume per unit time or milliliters per minute, for example.

25 Figure 12 shows a circular cylinder representing blood cells in a vessel moving at a particular velocity and thus reflecting energy at a specific Doppler frequency. The figure assumes that methods such as those in the referenced invention, for example, have been used to measure the 3-D orientation of the vessel so that the vector velocity can be calculated and the azimuth axis can be defined to be perpendicular to

30 the vessel.

The simplest way to estimate volume flow is to measure the vessel diameter, d , (or radius $d/2$), calculate the cross-sectional area, $A = \pi(d/2)^2$, and multiply by the average velocity. A more accurate way is to integrate the velocity as a function of position, over the cross-section. The velocity is a function of the radius, a , of the

35 ~~WO 00/72756~~ ~~PCT/US00/14691~~

cylinder depicted in Fig. 12. If a is the distance from the cylinder to its axis, and $v(a)$ is the velocity function, then the volume flow is

$$2\pi \int_0^{d/2} av(a)da \quad (7)$$

Equation (1) assumes a circular cross-section of constant radius, $r=d/2$. It is a
 5 special case of the more general polar coordinate integration:

$$\int_0^{2\pi} \left(\int_0^{r(\theta)} av(a, \theta) da \right) d\theta \quad (8)$$

The velocity function is determined by determining the diameter (and hence the
 radius) of the cylinder corresponding to each velocity.

For example, a 1.5-cm diameter Doppler ultrasound transducer array operating at 10
 10 MHz will be oriented with the length or azimuth direction perpendicular to the vessel
 to produce a B-scan (depth-azimuth) image. At a depth of approximately 10 mm, the
 cross range resolution is 0.1 mm. If the vessel diameter is 1 mm, the diameter can be
 measured with an accuracy of $\pm 5\%$. The area of the vessel is thus known to an
 accuracy of 10%. Since the average vector velocity can be measured extremely
 15 accurately, the volume flow is also accurate to $\pm 10\%$. The best accuracy is attained by
 measuring the azimuth extent corresponding to various velocities and then numerically
 evaluating equation (7) or (8). Naturally, a skilled artisan can readily program a
 processor to solve these equations, and calculate blood flow volume using routine
 programming techniques.

20 Since the autocorrelation function (pulse-to-pulse, at a fixed range) and the Doppler
 Power Spectrum form a Fourier pair, the total power can be obtained either as the
 autocorrelation function at zero lag or the integral of the Doppler Power Spectrum
 (Spectral Density) over all Doppler frequencies. Since radial velocity is proportional
 25 to Doppler frequency, the mean velocity can be obtained from the autocorrelation
 function as shown below:

$$R_{xx}(\tau) = \frac{1}{2\pi} \int_{-\infty}^{\infty} S(\omega) e^{j\omega\tau} d\omega = \int_{-\infty}^{\infty} S_d(f) e^{j2\pi f\tau} df$$

hence

$$R_{xx}(0) = \int_{-\infty}^{\infty} S_d(f) df = P_d = \text{total Doppler power}$$

and

$$R_{xx}(\tau) = R_{xx}(\tau) = \int_{-\infty}^{\infty} [j2\pi f S_d(f)] e^{j2\pi f\tau} df$$

leading to

$$R_{xx}(0) = j \int_{-\infty}^{\infty} [2\pi f S_d(f)] e^{j2\pi f\tau} df.$$

Thus

$$-j \frac{R_{xx}(0)}{R_{xx}(0)} = 2\pi \int_{-\infty}^{\infty} f \frac{S_d(f)}{\int_{-\infty}^{\infty} S_d(f) df} df = 2\pi E\{f_d\}$$

where the Doppler frequency and its mean (expected value) are related to the radial blood velocity and its mean by

$$f_d = \frac{2f_0}{c} v.$$

Hence

$$E\{v\} = \int_{-\infty}^{\infty} v \frac{P(v)}{\int_{-\infty}^{\infty} P(v) dv} dv = -j \frac{c}{4\pi f_0} \frac{R_{xx}(0_s)}{R_{xx}(0)}$$

10

which is used in the autocorrelation method of color-flow blood-velocity imaging. We note that if we do not normalize by dividing by the total Doppler power, we obtain a power-velocity product that indicates the volume flow rate. This is due to the fact that power is directly proportional to area [see Reference 1].

15

Since all velocity vectors are parallel at the narrowest point (the vena contracta), flow at that particular point can be considered as non-turbulent, even though severe turbulence exists before and after. Reference I shows that regurgitant blood flow

20

through the mitral heart valve can be quantitatively measured by observing the Doppler

20

-spectrum at that point and using the power-velocity-Integral relation below.

50

In terms of the velocity power spectrum, $P(v) = (2f_0/c) S_d(f)$, Reference I shows that the blood vessel area in a ~slice~ perpendicular to the line of sight is directly proportional to the total Doppler power (the total power at the output of the high-pass wall filter).

$$A = \frac{A_0}{P_0} P_d = \frac{A_0}{P_0} \int P(v) dv$$

where A_0 and P_0 are the known area and measured power in a narrow beam, smaller than the vessel. If the blood flow makes an angle θ with the line of sight, the area, and hence power, is increased by the factor $1/\cos\theta$. This offsets the fact that only the radial component of velocity is measured, so that the power velocity integral provides true volume flow:

$$Q = dQ/dt = \frac{A_0}{P_0} \int v P(v) dv$$

In Reference 1, P was measured with the same probe as P_0 by masking the outside of the aperture in order to create a wider beam. With our 2-D phased array, we would merely turn off or ignore some of the outer elements. More importantly, we can use the 3-D image to precisely locate the vena contracta, and lock on to it using monopulse. We can even monitor the valve during a stress test, while the patient is on a treadmill.

We note here, that there are several ways to measure the volume flow rate. Reference I uses the fact that it is proportional to the integral of the product of the velocity and the power per unit velocity, as in the last equation. An other way is to recognize that it is equal to the product of the average radial velocity and the total projected area that is, in turn, proportional to total Doppler power. Since the total Doppler power is used in the denominator of the autocorrelation-based color-flow velocity map, volume flow rate can be obtained by merely not dividing by the total power. If the P' pulse return (after MTI or Doppler high-pass or wall filtering) is

$$z_i = x_i + j y_i \quad i = 1, 2, \dots, N,$$

the volume flow rate is proportional to

51

$$\sum_{i=2}^N x_i y_{i-1} - y_i x_{i-1}$$

The normalization (denominator) that is used to convert this last quantity to mean
30 velocity can be

$$\sum_{i=2}^N x_i x_{i-1} + y_i y_{i-1}$$

35 that is based on a derivation in Reference 2, or a simple power estimate, such as

$$\sum_i x_i^2 + y_i^2$$

40 5

The point we wish to make here is that by not dividing by a power estimate to obtain
radial velocity, we obtain volume flow. Current ultrasound Doppler imaging systems
compute the mean velocity as a ratio,

$$E(v) = F / P_d,$$

and display it as a color flow image. Newer imaging systems [2] also display total
Doppler power (at the output of the wall fitter), P_d . By not dividing the color flow image
by P_d , we can also display the true volume flow, dQ/dt . This is because the
20-15 numerator,

$$F = P_d \cdot E(v),$$

is the power-velocity-integral that is directly proportional to the volume flow.

25 20

Determination of the scale factor, $A_0/P_0 = A/P_0 = dQ/dt/F$, that must multiply F to obtain
volume flow requires further comment.

A_0 is the area of a reference beam. In [1], A_0 is smaller than the blood flow area. We
30-25 will describe three normalization approaches.

1. Use a single transmit beam, wider than the vessel, and two simultaneous
receive beams. One receive beam (the measurement beam) is the same as the
transmit beam and the other (the reference beam) is smaller than the
35 30 vessel.

2. Use two (sequential or multiplexed) two-way (transmit and receive) beams. One (the measurement beam) is wider than the vessel and the other (the reference beam) is smaller than the vessel.
 3. Use two (sequential or multiplexed) two-way (transmit and receive) beams. Both are wider than the vessel and the measurement beam is wider than the reference beam.
- ~~5-35~~

Let A_0 be the known area of the reference beam, let P_0 and P_I be the measured received power in the reference and measurement beams. In case 1, the transmit

105 power density is the same for measurement and reference. The receive power is proportional to area. If the area of the vessel (in a slice perpendicular to the line of sight) is A , it follows that

$$A/A_0 = P_1/P_0.$$

1510 In cases 2 and 3, the transmit power density is greater in the reference beam than in the measurement beam, but by a known factor. In all three cases, the power received in the measurement beam is proportional to the vessel area. In case 2, the received reference power also varies with vessel size, but at a different rate than in the measurement beam. With proper calibration, correct measurements can be
2015 attained in all three cases.

[1]. T. Buck, Et al, "Flow Quantification in Valvular Heart Disease Based on the Integral of Backscattered Acoustic Power Using Doppler Ultrasound," *Proc. IEEE*, vol.88, no.3, pp.307-330, March 2000.

2520 [2]. K. Ferrara and G DeAngelis, "Color Flow Mapping", *Ultrasound in Medicine and Biology*, vol.23, no.2, pp.321-345, March 1997.

3025 EXAMPLE III 3-D DOPPLER ULTRASOUND BLOOD FLOW MONITOR
WITH ENHANCED FIELD AND SENSITIVITY

This example sets forth an ultrasound Doppler device and method that enables non-invasive diagnosis (the conventional role of ultrasound systems), and also non-invasive unattended and continuous monitoring of vascular blood flow for medical applications. In particular, the embodiment of the present invention set forth in this
3530 example provides: (1) affordable three-dimensional imaging of blood flow using a

low-profile easily-attached transducer pad, (2) real-time vector velocity, and (3) long-term unattended Doppler-ultrasound monitoring in spite of motion of the patient or pad. None of these three features are possible with current ultrasound equipment or technology.

35 The pad and associated processor collects and Doppler processes ultrasound blood velocity data in a three-dimensional region through the use of a two-dimensional phased array of piezoelectric elements on a planar, cylindrical, or spherical surface. Through use of unique beamforming and tracking techniques described herein, the present invention locks onto and tracks the points in three-dimensional space that
40 5 produce the locally maximum blood velocity signals. The integrated coordinates of points acquired by the accurate tracking process is used to form a three-dimensional map of blood vessels and provide a display that can be used to select multiple points of interest for expanded data collection and for long term continuous and unattended blood flow monitoring. The three dimensional map allows for the calculation of vector
45 10 velocity from measured radial Doppler.

A thinned array (greater than half-wavelength element spacing of the transducer array) is used to make a device of the present invention inexpensive and allow the pad to have a low profile (fewer connecting cables for a given spatial resolution). The array is thinned without reducing the receiver area by limiting the angular field of
20 15 view. Grating lobes due to array thinning can be reduced by using wide bandwidth and time delay steering. The array, or portions of the array, is used to sequentially insonate the beam positions. Once the region of interest has been imaged and coarsely mapped, the array is focused at a particular location on a particular blood vessel for measurement and tracking. Selection of the point or points to be
25 20 measured and tracked can be based on information obtained via mapping and may be user guided or fully automatic. Selection can be based, for example, on peak response within a range of Doppler frequencies at or near an approximate location.

In the tracking mode a few receiver beams are formed at a time: sum, azimuth difference, elevation difference, and perhaps, additional difference beams, at angles
30 25 other than azimuth (= 0 degrees) and elevation (= 90 degrees). Monopulse is applied at angles other than 0 and 90 degrees (for example 0, 45, 90, and 135 degrees) in order to locate a vessel in a direction perpendicular to the vessel. When the desired (i.e. peak) blood velocity signal is not in the output, this is instantly recognized (e.g., a monopulse ratio, formed after Doppler filtering, becomes non
35 30 zero) and the array is used to track (slow movement) or re-acquire (fast movement)

the desired signal. Re-acquisition is achieved by returning to step one to form and Doppler-process a plurality of beams in order to select the beam (and the time delay or “range gate”) with the most high-Doppler (high blood velocity) energy. This is followed by post-Doppler monopulse tracking to lock a beam and range gate on to
35 the exact location of the peak velocity signal. In applications such as transcranial
Doppler, where angular resolution based on wavelength and aperture size is inadequate, fine mapping is achieved, for example, by post-Doppler monopulse tracking each range
5 cell of each vessel, and recording the coordinates and monopulse-pair angle describing the location and orientation of the monopulse null.

10 With a three-dimensional map available, true vector velocity can be computed. For accurate vector flow measurement, the monopulse difference is computed in a direction orthogonal to the vessel by digitally rotating until a line in the azimuth-elevation or C-
10 scan display is parallel to the vessel being monitored. The aperture is more easily rotated in software (as opposed to physically rotating the transducer
15 array) if the aperture is approximately circular (or elliptical) rather than square (or rectangular). Also, lower sidelobes result by removing elements from the four corners of a square or rectangular array in order to make the array an octagon.

20 All currently available ultrasound devices (including “Doppler color flow mapping” systems) form images that are limited by their resolution. In some applications, such
20 as TCD, the low frequency required for penetration of the skull makes the azimuth and elevation resolution at the depths of interest larger than the vessel diameter. In this invention, as long as (1) a blood vessel or (2) a flow region of a given velocity can be
25 -resolved by finding a 3-D resolution cell through which only a single vessel passes, that vessel or flow component can then be very accurately located within the
25 cell. Monopulse is merely an example of one way to attain such sub-resolution accuracy (SRA). Other methods involve usuper..resolution~~ or “parametric” techniques used in “modern spectral estimation”, including the MUSIC algorithm and
25 autoregressive modeling, for example. SRA allows an extremely accurate map of 3-D flow.

30 This invention utilizes post-Doppler, sub-resolution tracking and mapping; it does Doppler processing first and uses only high Doppler-frequency data. This results in extended targets since the active vessels approximate “lines” as opposed to “points”. In
30 -three-dimensional space, these vessels are resolved, one from another. At a particular range, the monopulse angle axis can be rotated (in the azimuth-elevation

35- plane) so that the “line” becomes a “point” in the monopulse angle direction. That

point can then be located by using super-resolution techniques or by using a simple technique such as monopulse. By making many such measurements an accurate 3-D map
35 -of the blood vessels results.

Methods for extending the angular field of view of the thinned array (that is limited by
5- grating lobes) include (1) using multiple panels of transducers with multiplexed processing channels, (2) convex V-shaped transducer panels, (3) cylindrical shaped transducer panel, (4) spherical shaped transducer panel, or (5) negative ultrasound lens:

.

5 -If needed, moving the probe and correlating the sub-images can create a map of an even larger region.

40 Active digital beamforming can be utilized, but the implementation depends on a choice to be made between wideband and narrowband implementations. If emphasis is on high resolution mapping of the blood vessels, then a wide bandwidth (e.g., 50% of
10 -the nominal frequency) is used for fine range resolution. If emphasis is on Doppler spectral analysis, measurement, and monitoring, the map is only a tool.

45 In this case, a narrowband, low cost, low range-resolution, high sensitivity implementation might be preferred. A wideband implementation would benefit in performance (higher resolution, wider field of view, and reduced grating lobes) using time-delay steering while
15 -a narrowband implementation would benefit in cost using phase-shift steering. The invention can thus be described in terms of two preferred
20 implementations.

In a wideband implementation, time delay steering can be implemented digitally for
20 -both transmit and receive by over-sampling and digitally delaying in discrete sample intervals. In a narrowband implementation, (1) phase steering can be implemented digitally (digital beamforming) for both transmit and receive, and (2) bandpass
25 sampling (sampling at a rate lower than the signal frequency) can be employed with digital down-conversion and filtering.

25 Overview of this Embodiment.

This embodiment of the present invention involves (1) a family of ultrasound sensors, (2) the interplay of a set of core technologies that are unique by themselves, and (3)
30 a number of design options which represent different ways to implement the invention. To
30 -facilitate an organizational understanding of this many-faceted invention, a discussion of each of the three topics above follows.

The sensors addressed are all two-dimensional (i.e., planar or on the surface of a convex

shape such as a section of a cylinder) arrays of piezoelectric crystals for use

in active, non-invasive, instantaneous (or real-time), three-dimensional imaging and
-35 monitoring of blood flow. While the sensors and the techniques for their use apply to
all blood vessels in the body, the figures and detailed description emphasizes the
transcranial Doppler (TCD) monitor method as a nonlimiting example. The method
5 of the present invention utilizes a new, useful and unobvious approach to 3-D imaging
of blood velocity and blood flow that (1) allows for finer image resolution than would
otherwise be possible with the same hardware complexity (number of input cables and
associated electronics) and (2) allows for finer accuracy than would ordinarily be
possible based on the resolution. The invention measures and monitors
10 3-D vector velocity rather than merely the radial component of velocity.
The core technologies that constitute the invention are (1) array thinning with large
elements and limited scanning, (2) array shapes to reduce peak sidelobes and extend the
field of coverage, (3) post-Doppler sub-resolution tracking, (4) post-Doppler sub-
resolution mapping, (5) additional methods for maximizing the angular
15 field of view, and (6) various digital beamforming procedures for implementing the
mapping, tracking, and measurement processes. The invention encompasses array thinning,
where the separation between array elements is significantly larger than half the wavelength.
This reduces the number of input cables and input signals to be processed while maintaining
high resolution and sensitivity and avoiding ambiguities.
20 In the TCD application, where signal to noise and hence receiver array area is of
paramount importance, array thinning is possible without reducing the receiver array area
because a relatively small (compared to other applications) angular field of view is
needed.
Thinning with full aperture area imposes limitations on the angular field of view.
25 Methods for expanding the field of view include using more elements than are active
at any one time. For example, if the electronics is switched between two identical
panels, the cross-range field of view at any depth is increased by the size of the panel. If
the panels are pointed in slightly different directions so that overlapping or redundant
beams are avoided, the field of view is doubled. A generalization of this
30 approach involves the use of an array on a cylindrical or spherical surface.
In the TCD application, the achievable angular resolution is poor, regardless of the
method of thinning, or whether or not thinning is used. Once a section of a blood vessel is
resolved from other vessels in Doppler, depth, and two angles (az and el), Post-Doppler
sub-resolution processing locates that section to an accuracy that is
35 one-tenth to one-twentieth of the resolution. This allows for precise tracking and

accurate mapping. Tracking provides for the possibility of unattended long term monitoring and mapping aids the operator in selecting the point or points to be monitored.

- 5 One of ordinary skill in the art will readily recognize that there are many options
5- available in the design of any member of the family of sensors that utilizes any or all
of the core technologies that comprise this invention, all of which are encompassed by the
present invention. A two-dimensional array is established art that can be designed in
many ways and can have many sizes and shapes (rectangular, round, etc.).
- 10 As with other nonlimiting embodiments of the present invention set forth above, this
embodiment is a non-invasive, continuous, unattended, volumetric, blood vessel
tracking, ultrasound monitoring and diagnostic device for blood flow. It will enable
unattended and continuous blood velocity measurement and monitoring as well as 3-
dimensional vascular tracking and mapping using an easily attached, electronically
15 steered, transducer probe that can be in the form of a small pad for monitoring
application, when desired. Although a device of the present invention has
applications with blood vessels in any part of the body, the cranial application will
be used as a specific example. A device of the present invention can, for example:
1. Measure and continuously monitor blood velocity with a small low-profile
20 probe that can be adhered, lightly taped, strapped, banded, or otherwise easily attached
to the portion of the body where the vascular diagnosis or monitoring is required.
 2. Track and maintain focus on multiple desired blood vessels in spite of movement.
 - 25 3. Map 3-D blood flow; e.g., in the Circle of Willis (the central network of arteries that feeds
the brain) or other critical vessels in the cranial volume.
 4. Perform color velocity imaging and display a 3-D image of blood flow that is rotated via
track ball or joystick until a desired view is selected.
 5. Form and display a choice of projection, slice, or perspective views, including
30 (1) a projection on a depth-azimuth plane, a B-scan, or a downward-looking
perspective, (2) a projection on an azimuth-elevation plane, a C-scan, or a forward-
looking perspective, or (3) a projection on an arbitrary plane, an arbitrary slice, or an
arbitrary perspective.

6. Use a track ball and buttons to position circle markers on the points where measurement or monitoring of vector velocity is desired.
7. Move the track location along the blood vessel by using the track ball to slide the circle marker along the image of the vessel.

- 5 8. Display actual instantaneous and/or average vector velocity, estimated
average volume flow, and/or Doppler spectral distribution.
9. Maintain a multi-day history and display average blood velocity versus time
for each monitored vessel over many hours.
10. Sound an alarm when maximum or minimum velocity is exceeded or when
10 emboli count is high; and maintain a log of emboli detected.
11. Track, map, and monitor small vessels (e.g., 1mm in diameter), resolve
vessels as close as 4 mm apart (for example), and locate them with an accuracy of
 ± 0.1 mm, for example.

15 This embodiment of the present invention will allow a person with little training to
apply the sensor and position it based on an easily understood ultrasound image
display. The unique sensor can continuously monitor artery blood velocity and volume
flow for early detection of critical events. It will have an extremely low profile for
easy attachment, and can track selected vessels; e.g., the middle cerebral artery
(MCA), with no moving parts. If the sensor is pointed to the general volume location
20 of the desired blood vessel (e.g., within ± 1 cm.), it will lock to within ± 0.1 mm of the
point of maximum radial component of blood flow and remain locked in spite of
patient movement.

A device of the present invention can remain focused on the selected blood vessels
regardless of patient movement because it produces and digitally analyzes, in real
25 time, a 5-dimensional data base composed of signal-return amplitude as a function of:

1. Depth,
2. Azimuth,
3. Elevation,
- 30 4. Radial component of blood velocity,
5. Time.

Since a device of the present invention can automatically locate and lock onto the
35 point in three dimensions having the maximum high-Doppler energy, i.e., maximum
volume of blood having a significant radial velocity, unattended continuous blood
velocity monitoring is one of its uses. By using the precise relative location of the
5 point at which lock occurs as a function of depth, a device of the present invention
can map the network of blood vessels as a 3-dimensional track without the hardware
and computational complexity required to form a conventional ultrasound image.
Using the radial component of velocity along with the three-dimensional blood path,
a device of the present invention can directly compute parameters of blood flow, such
10 as vector velocity, blood flow volume, and Doppler spectral distribution.
A device having applications in a method of the present invention is a non-
mechanical Doppler ultrasound-imaging sensor comprising probes, processing
electronics, and display. Specific choices of probes allow the system to be used for
transcranial Doppler (TCD), cardiac, dialysis, and other applications.
15 Just as with other embodiments of the present invention set forth above, this
embodiment has application for medical evaluation and monitoring multiple locations
in the body. However, the transcranial Doppler application will be used as an
nonlimiting example. Fig. 13 shows the overall TCD configuration and a typical
definition of the display screen. The TCD system is comprised of one or two probes
20 that may be attached to the head with a "telephone operator's band" or a Velcro
strap. The interface and processing electronics is contained within a small sized
computer. A thin cable containing from 52 to 120 micro coax cables, depending on
the example probe design used, attaches the probe to the electronics in the
computer. When the operator positions the probe on the head and activates the
25 system, the Anterior, Middle and Posterior Cerebral Arteries and the Circle of Willis
are mapped on the screen along with other blood vessels. The arteries or vessels of
interest are selected by manually locating a cursor overlaid on the vessel 3-D map.
The system locks onto the blood vessels and tracks their position electronically. A
variety of selected parameters are displayed on the screen; e.g., the velocity, the
30 pulse rate, depth of region imaged, gain and power level. Using only one probe the
LCD can monitor multiple arteries (vessels) at a time. By way of example, presented
on the screen are dual traces, one for each artery selected. The blood velocity can be
dynamically monitored. As shown in Fig. 13 both the current blood velocity (dark
traces) and any historic trace (lighter color) can be displayed simultaneously. The
35 average blood velocity or estimated average flow for each artery is displayed below

the respective velocity trace. The image shows the arteries and the channel used for
-each artery. When two probes are used, the display is split showing signals from
both of them. For example, using a different probe (i.e., different size) with the same
electronics and display, the unit can be used to measure and monitor the blood flow
5 in a carotid artery. Similarly, it can be used to perform this function for dialysis,
5 anesthesia, and in other procedures.

The sensor is a two dimensional array of transducer elements (e.g., piezoelectric
crystals) that are electronically activated in both transmit and receive to effect a scan.
For example, if a square ($N \times N$) array is used, up to N^2 elements could be used at the
40 same time. This is illustrated in FIG. 14 for the case of $N=8$. The array need not be
10 square. Any $M \times N$ array may be utilized in this manner. All received signals (52 in the
example of FIG. 13) are sampled, digitized, and processed. This can be done, for
example, in a desk top or lap top personal computer with additional cards for
electronics and real-time signal processing as illustrated in FIG. 13 and FIG. 21. The
45 array is phase steered or time-delay steered, depending on the bandwidth utilized,
15 -which depends in turn on the desired range resolution. The angular field of view
shown in FIG. 15 is limited by the existence of grating lobes caused by array thinning
(spacing the array elements more than $\sim 1/2$ wavelength apart). The concept is
illustrated below for a 1-dimensional array forming a beam that measures only one
20 angle. For a two-dimensional array, this represents a horizontal or vertical cut
20 through the cluster of beams shown in FIG. 15.

The frequency utilized for TCD is usually at or near 2 MHz because higher
frequencies do not propagate well through bone and lower frequencies do not
provide adequate reflection from the blood cells. However, other frequencies have
25 applications when examining other parts of the body. With a propagation velocity of
25 1.54 millimeters per microsecond, the wavelength is 0.77 mm. If a filled array is
utilized, the element size and array pitch would be $d=0.77/2$. For a cross-range
resolution of 5.8 mm or less at a depth of 60 mm, the array size, L , must be at least 8
mm (Resolution = depth \times wavelength / L). Since $N=L/d$ in FIG. 2, N must exceed 21
30 and hence the array must have on the order of N^2 or over 400 elements. If the desired
30 -resolution is halved, the array size doubles and the number of elements exceeds
1,600. The array in FIG. 14 is said to be "thinned" because it only has 52 elements.
As explained above, "grating lobes" are ambiguities or extra, unwanted, beams
35- caused by using a transducer array whose elements are too large and hence too far

apart. The following analysis illustrates grating lobe suppression for the worst case of
 35 -narrowband signals and phase-shift beam processing. Time delay processing using
 wideband signals would be similar, but would further attenuate or eliminate grating
 lobes, resulting in even better performance. Naturally, one of ordinary skill in
 5 the art can readily program a processor to suppress or limit grating lobes with the
 equations described herein using routine programming techniques.

5

Let

$$x = (d/\lambda) \sin \theta, \quad (9)$$

represent a normalization for the *angle*, θ , from which reflected acoustic energy
 10 arrives. The azimuth (or elevation) angle, θ , is zero in the broadside direction,
 perpendicular to the transducer array and d is the width (or length) of a single
 element of the receiver array. The wavelength of the radiated acoustic wave is $\lambda =$
 c/f , where c is the acoustic propagation velocity (1540 meters/second in soft tissue)
 and f is the acoustic frequency (usually between 1 and 10 megahertz). The wide
 15 pattern in FIG. 16a is the element pattern

$$a_e(x) = \sin \pi x / \pi x. \quad (10)$$

The pattern is the product of the element pattern, the array pattern, and $\cos \theta$

20

$$a(\theta) = \cos(\theta) a_e(x) a_a(x) \quad (11)$$

Each of the two component patterns is plotted separately as a function of θ in FIG.
 16a and the total pattern of equation (11) is plotted in FIG 16b. In the far-field, i.e.,
 25 for $\lambda r \gg L^2$, where r is the range or depth and L is the length of the aperture, the array
 pattern steered to the angle $\theta = \theta_0$ is

$$a_a(x) = \sum_{n=0}^{N-1} w_n e^{j2\pi n(x-x_0)} \quad (12)$$

30 where w_n is a weighting to reduce sidelobes and N is the number of elements in one
 dimension. As seen in Figure 16a, equation (12) is periodic in x . The peak at $\theta = \theta_0$
 ($\theta_0 = 0$ in Fig 16) is the desired beam and the others are grating lobes.

In the near field, when focused at (r_0, θ_0) , equation (12) is replaced by the slightly
35 better general Fresnel approximation:

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$$a_a(x, z) = \sum_{n=0}^{N-1} w_n e^{j2\pi \left[n(x-x_0) + \left(n - \frac{N-1}{2} \right)^2 (z-z_0) \right]} \quad (13)$$

(provided that that the range significantly exceeds the array size, $r \gg L$), where $x = d \sin \theta / \lambda$, as before, and

$$z = d^2 \cos^2 \theta / \lambda r \quad (14)$$

Because the receiver aperture is sampled with a spatial period of d , the receiver array pattern will be periodic in $\sin \theta$, with a period of λ/d (equation 12). This periodicity means that the array pattern is ambiguous. When the array is pointed broadside ($\theta = 0$), it will also be pointed at the angle $\theta = \sin^{-1}(\lambda/d)$, for example. In terms of the normalized variable, x , the period is unity. Since $|\sin \theta|$ cannot exceed 1, the variable x is confined to the interval $[-\lambda/d, \lambda/d]$. The conventional element spacing is $d = \lambda/2$. Thus, in a conventional phased array, x is always between -0.5 and +0.5, and hence ambiguities are not encountered. In a highly thinned array ($d > \lambda$), there will normally be ambiguities or grating lobes as illustrated in FIG. 16a. The second grating lobe, at $x = 2$ or $\theta = \sin^{-1}(2\lambda/d)$, is not real when d does not exceed 2λ .

FIG. 16b shows that the unsteered total pattern does not exhibit grating lobes. In a two dimensional array, the elements could be rectangular instead of square ($dx \times dy$), and the results would still be valid. Similar results could be obtained for an array in which the elements are staggered from row to row (and/or column to column).

In FIG. 17 the same array is used as in FIG. 16, but the receiver element signals are combined with a phase taper that steers the beam to $x = 0.2$ or $\theta = 4.71^\circ$. In FIG. 17b, we see that the grating lobes are not completely suppressed, with the largest one at $x = -1 + 2 = -.8$ or $\theta = -19.18^\circ$. FIG. 18 shows this in decibels. The worst-case grating lobe is attenuated by at least 12 dB, even in the stressing case of extremely narrow band operation. These Figures were produced in MATLAB, using the following software (rn-file):

```
MPATTERN xpattern.x  Script to plot monostatic patterns vs. theta
Mt=90; wave_length = 0.77 ; d=1.875 , N=8,
k=d/wave_length
35  t = -Mt: 0.1:Mt;
    tr = pi. *t./180;
    x=k*sin(tr);
```

```
p=pi*x + eps; R=sin(p) ./p;
```

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```

R=R.*cos(tr);
n=0 : N-1;
    xo=0;
xo=0.2; steered
5  e=exp (j*n'*2*pi*(x-xo));
    w=nanning(N);
    E=(2/N)*w'*e;
E=(1/N)*ones(1,N)*e;
    subplot(211); plot (t,[abs(R);abs (E)]);
10 ER=abs(E).*abs(R); Monostatic
    subplot(212); plot (t,(abs(ER)));
    figure(2); plot (t, 20*log10(abs (ER)));
    zoom on;

```

15 The values of d and 2. used in the above example are representative for a transcranial Doppler application of the invention. If $f=2$ MHz is chosen for the center frequency, the wavelength is 0.77 mm. An 8x8 array with a width and/or length of $L=15$ mm, provides a one dimensional thinning ratio of $2d/2 = 4.87$. A 15 mm square array with half-wavelength elements would require more than 15,000

20 elements. By thinning, this number was reduced to 52 provided that the angular field of view is limited to $2 \times 4.719.42^\circ$. For a 1 cm array at 2 MHz, the hyperfocal distance (where the 3 dB focal region extends to infinity) is $L^2/42$ 3.25 cm. For a 15 mm array, the hyperfocal distance is 7.3 cm. Thus, a fixed focus probe suffices for this

25 focus in depth should be added to the linear phase distributions required to steer the beams.

Using the configuration described above, the cluster of beams *in* Fig. 15 is used to approximately locate the desired point for collecting the blood velocity signal. This is

30 done initially, and is repeated periodically, in "mapping dwells" that are interspersed with normal measurement dwells. For example the output of each beam in the cluster would be Doppler processed by performing an FF1 or equivalent transformation on a sequence of pulse returns. The pulse repetition frequency (PRF) would typically be less than or equal to 9 kHz to unambiguously achieve a depth of

35 8.5 cm for the TCD application. In order to obtain a velocity resolution liner than $\Delta v = 2$ cm per second (to distinguish brain death), a dwell of duration as long as $T = \lambda/(2\Delta v) = 20$ ms, or 170 pulses at 8.5 kHz, may be desired in the measurement mode.

During monostatic mapping, 21 beams are scanned. If a mapping dwell is to be

completed in 20 ms, only 8 pulses per beam are available, and an 8-pulse FF1 would 40 be utilized for each beam position.

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The example shown in Figs. 16 through 18 was an 8 by 8 receiver array forming a 5 by 5 cluster of beams. This is an example of an approximate rule of thumb for this invention, that an N element linear array is recommended for use in producing $N/2 + 1$ beams for N even and $[N+1]/2$ beams for N odd. Thus, a 16 by 10 element
5 rectangular array would preferably be used to form a 9 by 6 cluster of beams, though the actual number of beams formed is arbitrary.

Because receive beams are formed only in a limited angular region, a wide-angle receiver element pattern (which usually implies a small element) is not required. In
10 fact, the size of the receiver element can be as large as the element spacing. Thus the receiver array is “thinned” only in the sense that the element spacing exceeds a half wavelength. Since the element size also exceeds a half wavelength, the array area is not reduced. It is thinned only in terms of number of elements, not in terms of receiver area. Consequently, there is no reduction in signal-to-noise ratio, nor a
15 requirement for increased transmitter power.

FIG. 19 illustrates a means for increasing the angular field of view in the azimuth direction by extending the array horizontally. A similar scheme could be used vertically to extend the elevation F.O.V. The 52-element array of FIG. 14 becomes a
20 single panel of the extended array. One panel is active at a time in FIG. 19. The beamwidth for FIG. 14, in radians, is nominally given by λ/L . At a range or depth of R , the cross range resolution is $R\lambda/L$ (typically 3 to 5 mm). The F.O.V in millimeters at that same range is less than $N/2+1 = 5$ times that beamwidth. If a second panel is used in a planar configuration, the second panel translates the beam pattern to the
25 right (or left) by the width of the panel, $L=L_2/2$ (typically 8 mm). The field of view can be extended by more than this (it can even be doubled) by tilting the two panels in opposite directions to minimize the overlap in coverage of the two panels.

Figure 19, with $L_1 \approx L_2$, simultaneously provides: (1) a large F.O.V. in the L_2 direction
30 to allow for the simultaneous monitoring of two blood vessels more than an inch apart, (2) a large active array area for high sensitivity, and (3) a number of active elements below 60 and a total number of elements below 120. An alternative, shown in FIG. 20, has the array on the surface of a segment of a cylinder. This uses 52 elements at a time with a total of only 84 elements (and hence only 84 cables). The

35 $L_1 \times L_2'$ active array translates around the curved surface as the beam is scanned

horizontally. If a symmetric F.O.V. extension (azimuth and elevation) is desired, a spherical surface could be utilized.

Figure 21 is an overall block diagram depiction of the overall blood flow monitor.

5 Most functions are performed by means of software in the digital processor.

Naturally, one of ordinary skill in the art can readily program the processor to perform functions described herein using equations set forth herein and routine programming techniques. A possible implementation of the analog processing is diagrammed in FIG. 22. The ND converter can be a bank of converters or one or more converters multiplexed amongst the 52 channels. If an extended array such as shown in FIG. 10 19 or 20 were used, a switch would be included between the 52 processing channels in Figure 22 and the actual elements. Note that the 52 element array of FIG. 14 represents an 8x8 array with corners removed ($52=8 \times 8 - 4 \times 3$). Other possibilities include a 24 element array ($24 = 6 \times 6 - 4 \times 3$), a 120 element array ($120 = 12 \times 12 - 4 \times 6$), 15 etc.

The transmitter produces pulses for each active element at a pulse repetition frequency (PRF) of 8,500 pulses per second. Each pulse will be at a frequency of $f_0 = 2$ MHz and will have a bandwidth, B , of at least 250 kHz (e.g., a pulse no more than 4 microseconds long).

20 For measurement, only one or two beam positions need be insonated by a single probe. For mapping, many beam positions must be insonated, with several pulses on each for moving target indication (MTI) and/or Doppler processing. A measurement frame duration longer than 20 milliseconds (170 pulses at an 8.5 kHz 25 PRF) may not be necessary because of the non-stationary (pulsed) nature of human blood flow. Mapping, requires several (4 to 11) pulses per beam position and many (e.g., 21 to 36) beam positions per frame. Since the Doppler resolution for mapping is not as fine as in the measurement mode, longer mapping frames can be used. If only 21 beams are formed with 8 pulses on each or if up to 34 beams are formed 30 with only 5 pulses on each, a frame duration of 20 ms can be maintained even during search and mapping.

FIG. 22 shows 52 identical receiver chains comprising

1. Processor controlled time gain control and time gate (open for up to 26 35 microseconds for each pulse).

2. A limiter for dynamic range control.
3. A low noise amplifier (LNA).
4. A low pass filter (to pass $1 < f_0 + B/2$ (e.g., $I \sim < 2.125$ kHz) and reject $\sim f_j > 5.875$ kHz by at least 40 dB (assuming $f_0 = 2$ MHz and $B = 250$ kHz).

5

AID conversion (typically 12 to 16 bits) is performed at an 8 MHz rate for each channel in FIG. 22. This keeps the analog filtering requirements extremely simple. It also permits extremely large bandwidths (up to 2 MHz) and time-delay steering. For narrower bandwidths and phase-shift steering, bandpass analog filtering and much lower sampling rates (determined by B rather than f_0) could be used. For the 8 MHz sampling rate, either time-delay or phase-shift beam steering can be utilized (depending on signal bandwidth). FIG. 22 depicts time delay steering for the transmitter. The distance from each array element to each focal point (each beam center at a nominal depth (e.g., 60 mm for TCD) would be pre-computed and stored either as a time delay or as a phase shift (depending on the type of steering) for each element for each beam. If phase shift steering were utilized on transmit, the transmitted signal could be created digitally in the processor, followed by D/A conversion for each element. Hence Figure 22 represents only one possible embodiment of the invention.

20

An example of the digital receiver processing for the case of an 8 MHz sampling rate per channel is described below. The input is 208 12 or 16 bit samples per pulse (8 samples per microsecond x 26 microseconds to allow for a 4 cm deep radial mapping field of view), 8,500 pulses / second, and 52 channels. This results in a maximum average rate of $52 \times 208 \times 8500 = 91.9$ MegaSamples per second (or 1.84 million samples in a 20 ms frame). During measurement, the range interval can be narrowed to less than 1 cm, reducing the number of samples per pulse to 32. The average rate for measurement and monitoring becomes 14 megasamples per second. The receiver processing steps are as follows:

30

- **Buffer** (to allow subsequent processing to be performed at the average rate).
- **Digitally Down Convert to Baseband** (make I and Q). 52 channels in parallel. Multiply input samples by samples of a 2 MHz cosine wave and -sine wave to create In-phase and Quadrature samples, respectively. Since the samples are $\frac{1}{4}$ cycle apart, the multiplicands are all 0, 1, or -1, and hence no multiplications

are needed. If $r(j,p)$ is the real p^{th} sample from the f^{th} channel, the complex low-pass signal, $s(j,p)$, has a real part for $p = 0,1,2,3,4, 5,\dots$ given by

$$r(j,0), 0, -r(j,2), 0, r(j,4), 0,$$

and an imaginary part given by

$$0, -r(j,1), 0, r(j,3), 0, -r(j,5),$$

This provides a data rate 2 times the input rate because the data is now complex.

- **Pre-Decimation Low-Pass Digital Filter.** Filter 52 complex channels.

Pass $|f| < B/2$, reject $|f| > r - B/2$, where r is the sampling rate after sample rate decimation (e.g., 1 MHz). If $B = 250$ kHz, r could be as low as 500 kHz. If B is large, r could be 2 or 3 MHz. If receiver phase-shift steering were to be performed, the output samples would be computed at the decimated rate. If receiver time-delay steering is to be used, we output 8 million complex samples per second and postpone sample rate decimation until after beam formation.

- **Perform MTI or create coarse Doppler cells.** For every channel and every range sample, either digitally high-pass filter the sequence of pulse returns to suppress clutter from tissue and bone or perform 52×208 8-point discrete Fourier transforms (DFT's or FFT's) for each mapping frame. (Six points of the 8-point complex DFT provides 3 positive and 3 negative coarse Doppler cells.)

- **Perform Digital Beamforming.** Case 1: Time Delay Beamforming with Sample Rate Decimation uses a set of pre-computed time delays to reduce 52 complex channels with 208 samples per pulse to one of M (e.g. 21) complex beam outputs with 25 samples (range cells) per pulse. The example given here assumes 8:1 decimation.

The maximum delay is slightly less than $0.75 \mu = 6 T$, where $T = 1/8$ microsecond is the time between input samples. For a given pulse return, the k^{th} sample ($k = 1, 2, \dots, 25$) of the i^{th} beam, $i = 1, 2, \dots, M$, is denoted by $b(i, k)$. The p^{th} sample ($p = 1, 2, \dots, 208$) of the j^{th} input channel ($j = 1, 2, \dots, 52$) is denoted by $s(j,p)$. Let d_{ij} be the delay required for the signal in channel j to produce beam i .

For a given pulse return, the k^{th} complex 1 MHz rate output sample for beam i is

$$b(i,k) = \sum_{j=1}^{52} \{a_{ij}s(j,8[k+1]-b_{ij}) + (1-a_{ij})s(j,8[k+1]-b_{ij}-1)\}$$

where b_{ij} is the integer part of d_{ij}/T (between 0 and 6) and a_{ij} is the fractional remainder (between 0 and 1). Determine power or amplitude in each output Doppler bin as $1^2 + Q^2$ or its square root:

Case 2: Phase-shift beamforming of already decimated data involves only a

- 35 _____ sequence of inner products of 52-dimensional complex vectors of element values with a complex vector of representing the required phase shifts.
- Display Coarse Blood-Vessel Color- Flow Map. Coarse blood vessel map is
5 the set of range, azimuth, and elevation cells with high power, with 6 Doppler values. Blue and red represent positive and negative Doppler, with saturation related to radial velocity and intensity related to power.
 - Initialize Acquisition. The user, looking at an azimuth-elevation Coarse Map
10 (with depth automatically truncated to a set of values that should include the MCA), moves the transducer and looks for a high-intensity, saturated spot. He can center the probe on that spot or he can have a device of the present invention display a range interval corresponding to the ACA, in which case he can make sure that both vessels are well within the angular field of view of the probe.
 - 15 • Acquisition and Tracking of one or two points being monitored. This is done with a single transmit beam focused on the spot identified above for several frames. Digital Down-conversion, low-pass filtering, and MTI are performed as before, but beamforming is different. Five receive beams are simultaneously formed. These are a sum beam and four monopulse difference beams, all
20 steered to the same point as the transmit beam. Each monopulse beam is equivalent to the difference between the outputs of a pair of beams displaced on opposite sides of the focal point. The four monopulse pairs are in 45 degree intervals with the first being horizontal, and the third being vertical. The monopulse-difference output with the largest magnitude is divided by the output
25 of the sum beam. The imaginary part is the "monopulse ratio" used to re-steer the beam (in the difference pair direction) so that it is better centered on the vessel. This procedure can be repeated in an effort to drive all four monopulse ratios to zero.
 - Measurement and Tracking. Tracking continues as described above during the
30 measurement mode. Measurement is made with fine Doppler resolution (128 point FFT) applied to only the sum beam. In a 15 ms frame, data from 128 pulses are collected (52 channels, 6 range samples). The pulses are Hamming weighted and FFT'd. This produces 128 Doppler bins (for each range bin and element), 66.4 times a second. Real sum beam outputs would then be produced
35 (using monopulse-guided steering) for each of 64 to 126 of these Doppler bins.

- Track maintenance and re-acquisition. Tracking is continued in parallel with measurement. If a monopulse ratio suddenly deviates far from zero and is not brought back to zero in one or two iterations, loss of track is declared. Re-acquisition is attempted autonomously by re-steering the beam by an amount
5 determined by correlating a current color flow map with a stored earlier version (from before track was lost). If this is unsuccessful, (monopulse ratios do not all converge to zero) an alarm is sounded so that the user can return to repeat initialization of acquisition.

Correlation with previous maps will be periodically applied to prevent wandering of
10 the data collection point along the vessel being tracked.

For tracking purposes, a monopulse tracking method described above can be used.

FIG. 23 illustrates the segment of a vessel in a single resolution cell, after rotation.
15 The resolution cell shown is not a cube because the range resolution might be finer than the cross-range resolution. The illustrated circular cylinder represents blood cells in a vessel reflecting energy at a fixed Doppler frequency. These represent a cylindrical annulus of blood cells, at a constant distance from the vessel wall, moving with approximately the same velocity. In the single resolution cell of Fig. 23, the
20 return at the highest Doppler would represent a line in three-dimensional space (the axis of the vessel) and hence a point on the azimuth axis after rotation. When applied to the highest Doppler output, the Sum beam would have broad peak at zero azimuth ($\alpha = 0$) and the monopulse ratio, $r = Az/Sum$, will be a linear function of the azimuth angle to which the array is phase steered:

$$25 \quad r(\alpha) = k\alpha.$$

This result can be attained by applying the same phase across the aperture for the *Az* and *Sum* beams, but using the derivative of the *Sum* beam amplitude weights with respect to x and y respectively for the *Az* and *El* aperture weights.

30 Figures 24A-24D represent four different excitations of one transmitter transducer element of an ultrasonic probe. Each square wave (a, b, c, d) represents a constant amplitude pulse of ultrasonic excitation of defined pulse width. The series of five pulses illustrated in Fig. 24A will transmit a burst of acoustic energy into the patient at a defined power level. It has been found that, in order to reduce the level of side lobe transmission, it is desirable to excite probe elements at the edges of the probe (See, for

example, Fig. 2) at a lower power level than elements at the center of the probe. In the prior art this is done by reducing the amplitude of transducer element excitation, as illustrated in Fig. 24B.

5 In the inventive method described herein, the transmitted power is reduced by reducing the pulse width, as illustrated in Fig. 24C or by reducing the number of pulses in each burst, as illustrated in Fig. 24D. The reduced number of pulses can be at a reduced probe width to achieve a wide range of power level control, as illustrated in Fig. 24D, or at the full pulse width, which may be easier to implement electronically.

10 The amount of power reduction at the probe periphery should be sufficient to produce an operationally significant reduction of harmonics in the modulated signal.

Figure 25 illustrates exemplary circuitry used to excite probe elements at variable
15 amplitude and pulse count. Figure 26 illustrates the much simpler circuitry with which probe elements are excited at constant amplitude with variable pulse width and variable number of pulses.

The disclosure that follows illustrates the exemplary use of Doppler shifted acoustic
20 imaging to map and track blood flow. It illustrates the techniques and systems that profit from application of the hereindisclosed invention. Alternative probe geometries to which the hereindisclosed invention could be profitably applied are illustrated in US Pat. #6,524,253 B1, which is incorporated by reference in its entirety herein.

30 Many other variations and modifications of the invention will be apparent to those skilled in the art without departing from the spirit and scope of the invention. The above-described embodiments are, therefore, intended to be merely exemplary, and all such variations and modifications are included within the scope of the invention as defined in the appended claims.

CLAIMS

What is claimed is:

1. A method for transmitting a plurality of short series of bursts of ultrasonic energy pulses from a probe terminated by an array of transducer elements; and modulating an amount of energy transmitted by each burst from transducer element to transducer element across the probe by varying the width of each energy pulse or varying the number of pulses per burst or varying both the width of each pulse and the number of pulses per burst while keeping each pulse amplitude constant from pulse to pulse across the probe, the modulating being sufficient to produce an operationally significant reduction of harmonics in the modulated signal.
2. The method of Claim 1 in which the amount of energy transmitted by each burst is less at the array's edge than at the array's center.
3. The method of Claim 2 in which the amount of energy transmitted by each burst is progressively reduced from a maximum at the array's center to a minimum at the array's edge.
4. The method of Claim 2 in which the array is a one dimensional array or a two dimensional array.
5. An apparatus for transmitting a plurality of short series of bursts of ultrasonic energy pulses from a probe terminated by a one dimensional or a two dimensional array of transducer elements and modulating an amount of energy transmitted by each burst from

transducer element to transducer element across the probe, comprising control circuitry incorporating digital logic adapted for varying the width of each pulse or the number of pulses per burst or both the width of each pulse and the number of pulses per burst from burst to burst across the probe, the modulating being sufficient to produce an operationally significant reduction of harmonics in the modulated signal.

6. An apparatus of Claim 5 in which the amount of energy transmitted by each burst is less at the array's edge than at the array's center.
7. An apparatus of Claim 6 in which the amount of energy transmitted by each burst is progressively reduced from a maximum at the array's center to a minimum at the array's edge.

ABSTRACT

In traditional ultrasound applications the transmit signal is a short series of constant amplitude pulses. The pulses are 50 percent duty cycle and constitute a single frequency. The amplitude must be very consistent or system performance will suffer. Modern ultrasound requires shaping the transmitter beam by applying different amplitudes to an array of elements. However, the need to change voltage from one series to the next can cause problems with the electronics associated with the transmission. One problem is there is not enough time for the voltage to settle from one series to the next. This has the effect of causing artifacts in the image. One means of overcoming the problems is to hold the voltage constant and modify the modulation of the series of pulses to achieve the different power levels.

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